## Video chapters

Chapter-1 (Boolean Algebra \& Logic Gates):Introduction to Digital Electronics, Advantage of Digital System, Boolean Algebra, Laws, Not, OR, AND, NOR, NAND, EX-OR, EX-NOR, AND-OR, OR-AND, Universal Gate Functionally Complete Function.
-Chapter-2 (Boolean Expressions): Boolean Expressions, SOP(Sum of Product), SOP Canonical Form, POS(Product of Sum), POS Canonical Form, No of Functions Possible, Complementation, Duality, Simplification of Boolean Expression, K-map, Quine Mc-CluskyMethod.

- Chapter-3 (Combinational Circuits):Basics, Design Procedure, Half Adder, Half subtractor, Full Adder, Full Subtractor, Four-bit parallel binary adder / Ripple adder, Look ahead carry adder, Four-bit ripple adder/subtractor, Multiplexer, Demultiplexer, Decoder, Encoder, Priority Encoder
-Chapter-4 (Sequential Circuits):Basics,NOR Latch, NAND Latch, SR flip flop, JK flip flop, T(Toggle) flip flop, D flip flop, Flip Flops Conversion, Basics of counters, Finding Counting Sequence Synchronous Counters, Designing Synchronous Counters, Asynchronous/Ripple Counter, Registers, Serial In-Serial Out (SISO), Serial-In Parallel-Out shift Register (SIPO), Parallel-in Serial-Out Shift Register (PISO), Parallel-In Parallel-Out Shift Register (PIPO), Ring Counter, Johnson Counter
-Chapter-5 (Number Sysem\& Representations): Basics, Conversion, Signed number Representation, Signed Magnitude, 1's-Complement, 2's Complement, Graycode, Binary-Coded Decimal Code (BCD), Excess-3 Code.


## Basics

- Electrical engineering is a professional engineering discipline that generally, deals with the study and application of electricity, electronics and electromagnetism and heavy voltage devices like transformers, motors etc.

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- Electronic engineering is an electrical engineering discipline where we work on low voltage devices (such as semiconductor devices, especially transistor, diodes and integrated circuits) to design electronic circuits, VLSI devices and systems.

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- Electronic systems are generally of two types -
- Analog system in an analog representation, a continuous value is used to denote the information. Analog signal is defined as any physical quantity which varies continuously with respect to time. e.g. amplifier, cro, ecg. Watch, radio.



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- Digital system - the information is denoted by a finite sequence of discrete value or digits. Digital signal is defined as any physical quantity having discrete values. E.g. digital watch, calculator, bp machine, thermometer etc.



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## Advantage of Digital system

- Easy to Design - Digital systems are simpler to design because you mostly need knowledge of switching circuits. You don't need to be an expert in engineering or math.
- Long-Term Storage and Easy Processing - You can store information for a very long time, and it's easier to process data in digital systems.
- Higher Accuracy and Precision - Digital devices give more accurate and precise results. They are less affected by things like noise or electric fields.
- Better Modularity and Fabrication - Digital systems are modular, meaning they are built in parts that can be easily replaced or upgraded. They are also smaller in size, thanks to things like integrated circuits.

Cost-Effective - Generally, digital systems are less expensive to build and maintain.

Idempotent Law

$$
\begin{aligned}
& a \cdot a=a \\
& a+a=a
\end{aligned}
$$

## Associative law

a. (b. c) $=(\mathrm{a} . \mathrm{b}) . \mathrm{c}$
$a+(b+c)=(a+b)+c$

## Commutative law

a. $b=b . a$ $a+b=b+a$

## Distributive law

a. $(b+c)=a \cdot b+a . c$
$a+(b \cdot c)=(a+b) \cdot(a+c)$

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Identity law

$$
\begin{aligned}
& a+0=a \\
& a .0=0 \\
& a+1=1 \\
& \text { a. } 1=a
\end{aligned}
$$

Complementation law

$$
\begin{align*}
& 0^{\prime}=1 \\
& 1^{\prime}=0 \\
& \text { a. } a^{\prime}=0 \\
& a+a^{\prime}=1
\end{align*}
$$

Involution law

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$$
a+a^{\prime}=1
$$

LI
-

$$
\begin{aligned}
& \text { De-Morgan law } \\
& \begin{array}{l}
(a+b)^{\prime}=a^{\prime} \cdot b^{\prime} \\
(a \cdot b)^{\prime}=a^{\prime}+b^{\prime}
\end{array}
\end{aligned}
$$

Inerlicy

$$
5
$$


ation law
$\square$

$$
1
$$

$\square$ $\square$

$$
\left(a^{\prime}\right)^{\prime}=a
$$

Absorption law

- $a+a b=a$
- $a .(a+b)=a$

Compensation theorem

- $a b+a^{\prime} c+b c=a b+a^{\prime} c$
- $(a+b)\left(a^{\prime}+c\right)(b+c)=(a+b)\left(a^{\prime}+c\right)$
- $a+a^{\prime} b=a+b$
- $a \cdot\left(a^{\prime}+b\right)=a \cdot b$


## Not Gate(Inverter)

- It represents not logical operator is also known as inverter, it is a unary operator, which simply complement the input.



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## OR Gate

- It is a digital logic gate, that implements logical disjunction. The output will be high if at least one of the input lines is high.


| Truth Table |  |  |
| :---: | :---: | :---: |
| Input |  | Output |
| $\mathbf{A}$ | $\mathbf{B}$ | $\mathrm{Y}=\mathrm{A}+\mathrm{B}$ |
| 0 | 0 |  |
| 0 | 1 |  |
| 1 | 0 |  |
| 1 | 1 |  |

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## And Gate

- It is a digital logic gate, that implements logical conjunction. Output will be high if and only if all input are high otherwise low.


| Truth Table |  |  |
| :---: | :---: | :---: |
| Input |  | Output |
| A | B | $\mathrm{Y}=\mathrm{A} \cdot \mathrm{B}$ |
| 0 | 0 |  |
| 0 | 1 |  |
| 1 | 0 |  |
| 1 | 1 |  |

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## Nor gate

- The output will be high if and only if all inputs are low. Or simply a OR gate followed by an invertor.
- NOR gate is also called universal gate because it can be used to implement any other logic gate.


| Truth Table |  |  |
| :---: | :---: | :---: |
| Input |  | Output |
| $\mathbf{A}$ | B | $\mathrm{Y}=(\mathrm{A}+\mathrm{B})^{\prime}$ |
| 0 | 0 |  |
| 0 | 1 |  |
| 1 | 0 |  |
| 1 | 1 |  |

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## NAND Gate

- The output will be low if and only if all inputs are high. Or simply an and gate followed byan invertor
- NAND gate is also called universal gate.


| Truth Table |  |  |
| :---: | :---: | :---: |
| Input |  | Output |
| A | B | $\mathrm{Y}=(\mathrm{A} \cdot \mathrm{B})^{\prime}$ |
| 0 | 0 |  |
| 0 | 1 |  |
| 1 | 0 |  |
| 1 | 1 |  |

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## EX-OR

- For two inputs, output will be high if and only if both the input values are different.
- $a \oplus b=a^{\prime} \cdot b+a . b^{\prime}$


| Truth Table |  |  |
| :---: | :---: | :---: |
| Input |  | Output |
| $\mathbf{A}$ | B | $\mathrm{Y}=\mathrm{A} \oplus \mathrm{B}$ |
| 0 | 0 |  |
| 0 | 1 |  |
| 1 | 0 |  |
| 1 | 1 |  |

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## EX-OR

- The XOR gate is a digital logic gate that gives High as output when the number of inputs High are odd.


| $A$ | $B$ | $C$ | $a \oplus b \oplus c$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 |  |
| 0 | 0 | 1 |  |
| 0 | 1 | 0 |  |
| 0 | 1 | 1 |  |
| 1 | 0 | 0 |  |
| 1 | 0 | 1 |  |
| 1 | 1 | 0 |  |
| 1 | 1 | 1 |  |

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## EX-NOR

- For two input, output will be high if and only if both the input values are same - $\quad \mathrm{a} \odot \mathrm{b}=\mathrm{a}^{\prime} \cdot \mathrm{b}^{\prime}+\mathrm{a} . \mathrm{b}$


| Truth Table |  |  |
| :---: | :---: | :---: |
| Input |  | Output |
| $\mathbf{A}$ | $\mathbf{B}$ | $\mathbf{Y}=\mathbf{A} \odot$ |
| 0 | 0 |  |
| 0 | 1 |  |
| 1 | 0 |  |
| 1 | 1 |  |

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## EX-NOR

- The EX-NOR gate is a digital logic gate that gives output High when the number inputs low are even.


| $\mathbf{A}$ | $\mathbf{B}$ | $\mathbf{C}$ | $\mathrm{a} \odot \mathrm{b} \odot \mathrm{c}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 |  |
| 0 | 0 | 1 |  |
| 0 | 1 | 0 |  |
| 0 | 1 | 1 |  |
| 1 | 0 | 0 |  |
| 1 | 0 | 1 |  |
| 1 | 1 | 0 |  |
| 1 | 1 | 1 |  |
|  |  |  |  |

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## AND-OR(NAND-NAND) Implementation

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## OR-AND(NOR-NOR) Implementation

Implementing Every Gate with NAND Gate
NAND Gate Symbol
NOT Gate



AND Gate
Buffer

$A O-\bar{A} \quad$ OR Gate


|  | NOT | AND | OR | NAND | NOR | EX-OR | EX-NOR |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| NAND | 1 | 2 | 3 | 1 | 4 | 4 | 5 |

## Implementing Every Gate with NOR Gate



## Functionally Complete Function

- As we know there are only three fundamental Boolean operator NOT, AND and OR. All the other operators are derived from these operators
- We understand NOT along with OR(NOR) and NOT along with AND(NAND) are used as universal gates.
- Any Boolean operation or function can be implemented with NAND or NOR operation.


## $F=x^{\prime} y+x y^{\prime}+z$

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$$
F=b d+b c+a c d^{\prime}
$$

$$
F=\left(x^{\prime}+y\right)\left(x+y^{\prime}\right)(z)
$$

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## Boolean Expressions

- Boolean expressions are the method using which we save the information about the Boolean function, that when we get value 1 and when we get value 0 as output.
- So, we convert the truth table of the function into an expression, reading which we can understand when the output is 1 and when it is zero.

| Light | Day | Engine | Warning |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 0 | 1 |
| 1 | 1 | 1 | 1 |

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- There are two popular approaches for writing these expressions.
- Sum of Product (SOP) (which remember when we get 1)
- Product of Sum (POS) (which remember when we get 0)
- Make a note of this as we are studying Boolean function remembering both 0 and 1 is not required, so we can either concentrate on 0 or 1.

| Light | Day | Engine | Warning |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 0 | 1 |
| 1 | 1 | 1 | 1 |

- $W(L, D, E)=\Sigma_{m}(1,4,6,7)$

$$
\text { -W } W(L, D, E)=\Pi_{M}(0,2,3,5)
$$

- Power of SOP and POS bother are same, i.e. any Boolean functions can be represented using both SOP and POS.
- Mathematically speaking we should choose ( 0 or 1 ), whatever is less in number because then it will be easy to represent

| $A$ | $B$ | $f_{1}=$ | $f_{2}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 1 | 1 |

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## SOP (Sum of Product)

- A sum of product form expression contains product terms (AND terms) which are sum (OR) together, that's why called sum of product.
- Each product terms (AND terms) consists of one or more literals (variables) appearing either in complements or uncomplemented form. E.g. a'b + b'c' + abc
- A product term which contains all the literals (variables) either in complemented or uncomplemented form is called minterm.
- In a $n$ variable function, there will be $2^{n}$ minterms.

| Binary Representation | Sequence | Minterm | Designation |
| :---: | :---: | :---: | :---: |
| 000 | 0 | $a^{\prime} b^{\prime} c^{\prime}$ | $m_{0}$ |
| 001 | 1 | $a^{\prime} b^{\prime} c$ | $m_{1}$ |
| 010 | 2 | $a^{\prime} b^{\prime}$ | $m_{2}$ |
| 011 | 3 | $a^{\prime} b c$ | $m_{3}$ |
| 100 | 4 | $a^{\prime} b^{\prime}$ | $m_{4}$ |
| 101 | 5 | $a b^{\prime} c$ | $m_{5}$ |
| 110 | 6 | $a b c^{\prime}$ | $m_{6}$ |
| 111 | 7 | $a b c$ | $m_{7}$ |

- The result of a product term must always be 1 , If a literal is having value 1 then it is ok, but if not, then we complement those which is 0 , it to make it 1.
- There is only 1 input sequence of variables for any minterm on which the output is 1 , so it represents information.
- Then the sum of all product term(minterm) to from a function, and functions will have value 1, if at least one of the product term(minterm) is 1.

| Binary Representation | Sequence | M'aterm | Designation |
| :---: | :---: | :---: | :---: |
| 000 | 0 | $a^{\prime} b^{\prime} c^{\prime}$ | $m_{0}$ |
| 001 | 1 | $a^{\prime} b^{\prime} c$ | $m_{1}$ |
| 010 | 2 | $a^{\prime} b c^{\prime}$ | $m_{2}$ |
| 011 | 3 | $a^{\prime} b c$ | $m_{3}$ |
| 100 | 4 | $a b^{\prime} c^{\prime}$ | $m_{4}$ |
| 101 | 5 | $a b^{\prime} c$ | $m_{5}$ |
| 110 | 6 | $a b c^{\prime}$ | $m_{6}$ |
| 111 | 7 | $A b c$ | $m_{7}$ |
|  |  |  |  |

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## Canonical logic forms

- Either in POS or SOP form it is not essential that all product or sum terms contains all the literals.
- Canonical SOP form: - In a sum of product form expression, if each AND term (product term) consists all the literals(variables) appearing either in complements or uncomplemented form. E.g. $a^{\prime} b c+a b^{\prime} c^{\prime}+a b c$. Then the form is said to be canonical SOP.
- $F(a, b, c)=a+a^{\prime} b+a b c+b c^{\prime}$


## POS (Product of Sum)

- A product of sum (POS) form of expression contains OR (sum) terms which are AND (product) together, that's why called product of sum expression.
- Each OR term (sum term) consists of one or more literals(variables) appearing either in complemented or uncomplemented form. $\left(a^{\prime}+b\right) .\left(b^{\prime}+c^{\prime}\right) .(a+c)$
- A OR (sum) term which contains all the literals(variables) either in complemented or uncomplemented form is called maxterm. In a $n$ variable function, there will be $2^{n}$ maxterms.

| Binary Representation | Sequence | Maxterm | Designation |
| :---: | :---: | :---: | :---: |
| 000 | 0 | $a+b+c$ | $M_{0}$ |
| 001 | 1 | $a+b+c^{\prime}$ | $M_{1}$ |
| 010 | 2 | $a+b^{\prime}+c$ | $M_{2}$ |
| 011 | 3 | $a+b^{\prime}+c^{\prime}$ | $M_{3}$ |
| 100 | 4 | $a^{\prime}+b+c$ | $M_{4}$ |
| 101 | 5 | $a^{\prime}+b+c^{\prime}$ | $M_{5}$ |
| 110 | 6 | $a^{\prime}+b^{\prime}+c$ | $M_{6}$ |
| 111 | 7 | $a^{\prime}+b^{\prime}+c^{\prime}$ | $M_{7}$ |

- The result of the OR (sum) term must be 0 , so If a variable is having value 0 then it is ok, but if not then we complement the variable it to make it 0 .
- There is only 1 input sequence for which the output of a Maxterm is 0 . Because, it requires ali values as zero.
- Then the product of all sum term (maxterm), from a function, and functions will have a value 0 , if any of the sum term(maxterm) is 0 .

| Binary Representation | Sequence | Maxterm | Designation |
| :---: | :---: | :---: | :---: |
| 000 | 0 | $a+b+c$ | $M_{0}$ |
| 001 | 1 | $a+b+c^{\prime}$ | $M_{1}$ |
| 010 | 2 | $a+b^{\prime}+c$ | $M_{2}$ |
| 011 | 3 | $a+b^{\prime}+c^{\prime}$ | $M_{3}$ |
| 100 | 4 | $a^{\prime}+b+c$ | $M_{4}$ |
| 101 | 5 | $a^{\prime}+b+c^{\prime}$ | $M_{5}$ |
| 110 | 6 | $a^{\prime}+b^{\prime}+c$ | $M_{6}$ |
| 111 | 7 | $a^{\prime}+b^{\prime}+c^{\prime}$ | $M_{7}$ |

## Canonical logic forms

- Canonical POS form: - In a product of sum form expression, if each OR term (sum term) consists all the literals(variables) appearing either in complements or uncomplemented form. E.g. $\left(a^{\prime}+b+c\right) .\left(a+b^{\prime}+c^{\prime}\right) .(a+b+c)$. Then the form is said to be Canonical POS form.


## No of functions possible

Q With n-Boolean variables how many different Boolean functions are possible?
With $n$ binary variable we can generate $2^{n}$ combinations. And as we know that a Boolean function can also have only 2 possible values either 0 or 1 , so total number of different functions possible will be $2^{\wedge}\left(2^{n}\right)$.
Similarly, we can generalize this idea as $x^{\wedge}\left(y^{2}\right)$ are the total number of functions possible, $x$ is the nature of the function, $y$ is the nature of the variable and $z$ is the number of variables.

| $A$ | $B$ | $\mathbf{f}_{0}$ | $\mathbf{f}_{1}$ | $\mathbf{f}_{2}$ | $\mathbf{f}_{3}$ | $\mathbf{f}_{4}$ | $\mathbf{f}_{5}$ | $\mathbf{f}_{6}$ | $\mathbf{f}_{7}$ | $\mathbf{f}_{8}$ | $\mathbf{f}_{9}$ | $\mathbf{f}_{10}$ | $\mathbf{f}_{11}$ | $\mathbf{f}_{12}$ | $\mathbf{f}_{13}$ | $\mathbf{f}_{14}$ | $\mathbf{f}_{15}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| 0 | 1 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |
| 1 | 0 | 0 | 0 | 1 | 1 | 0 | 0 | 1 | 1 | 0 | 0 | 1 | 1 | 0 | 0 | 1 | 1 |
| 1 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 |

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## Complementation

- Let us consider a function $f(a, b, c, d, \ldots . . ., 0,1,+, \cdot)$, then the complement of the function is defined as $\quad f^{\prime}\left(a^{\prime}, b^{\prime}, c^{\prime}, \ldots . . ., z^{\prime}, 1,0, .,+\right)$.
- i.e. When all the variables are replaced by their compliments, $0 \rightarrow 1,1 \rightarrow 0$, or $\rightarrow$ and, and $\rightarrow$ or, then we will be called them compliment of a function.


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## Duality

- Let us consider a function $f(a, b, c, d, \ldots . .,, z, 0,1,+, \cdot)$, then the dual of the function is defined as $f^{d}(a, b, c, \ldots . . ., z, 1,0, \cdot,+)$.
- When the nature of variable remains same but $0 \rightarrow 1,1 \geqslant 0$, or $\rightarrow$ and, and $\rightarrow$ or, then they are called dual functions.

| $a$ | $b$ |  |  |
| :---: | :---: | :---: | :---: |
| $L 0 H$ | $L 0 H$ | L0H | L0H |
| LOH | $H 1 L$ | $H 1 L$ | L0H |
| $H 1 L$ | L0H | H1L | L0H |
| $H 1 L$ | $H 1 L$ | $H 1 L$ | H1L |

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- When we take dual of a function, then the functionality of a function remains the same but a positive logic system is transformed to negative logic system.
- If a function works corrects in positive logic system, then it must also work correct in negative logic system as it do not depends on magnitude.

| OR |
| :---: |
| NOR |
| $\mathrm{EX}-\mathrm{OR}$ |

dual

## Simplification of Boolean expression

- After deriving a Boolean expression from the truth table next important step is to minimise it, so that the cost of implementation into hardware and delay because of additional gates can be reduced. There are following methods for simplifying a Boolean expression
- Using Karnaugh-map
- Using Quine Mc-Clusky Method(Tabular Method)
- Using Algebraic method (Boolean Laws)
- Maurice Karnaugh (4 October 1924-8 Nov 2022) is an American physicist, mathematician and inventor known for the Karnaugh map used in Boolean algebra.
- Karnaugh map is one of the most extensively used tool, it is a graphical representation, represents truth table by pictorial form, provides a systematic method for simplifying or minimizing a Boolean expression. For a $n$-variable $k$-map, there will be $2^{n}$ cells addressed by a gray code. Each cell corresponds to one minterm or maxterm.

|  | a | b | c | d | F |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 0 | 0 | 0 | 0 |  |
| 1 | 0 | 0 | 0 | 1 |  |
| 2 | 0 | 0 | 1 | 0 |  |
| 3 | 0 | 0 | 1 | 1 |  |
| 4 | 0 | 1 | 0 | 0 |  |
| 5 | 0 | 1 | 0 | 1 |  |
| 6 | 0 | 1 | 1 | 0 |  |
| 7 | 0 | 1 | 1 | 1 |  |
| 8 | 1 | 0 | 0 | 0 |  |
| 9 | 1 | 0 | 0 | 1 |  |
| 10 | 1 | 0 | 1 | 0 |  |
| 11 | 1 | 0 | 1 | 1 |  |
| 12 | 1 | 1 | 0 | 0 |  |
| 13 | 1 | 1 | 0 | 1 |  |
| 14 | 1 | 1 | 1 | 0 |  |
| 15 | 1 | 1 | 1 | 1 |  |
|  |  |  |  |  |  |



|  | ab | $a^{\prime} b^{\prime}$ | a'b | ab | ab' |  | cd | $c^{\prime} d^{\prime}$ | c'd | cd | cd |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| cd |  | 00 | 01 | 11 | 10 | ab |  | 00 | 01 |  |  |
| $\mathrm{c}^{\prime} \mathrm{d}^{\prime}$ | 00 | - | 4 | 12 |  | $a^{\prime} b^{\prime}$ | 00 |  |  |  |  |
| c'd | 01 | 1 |  | 13 |  | a'b | 01 |  |  |  |  |
| cd | 11 | 3 | 7 | 15 |  | ab | 11 |  | 13 | 15 |  |
| cd' | 10 |  |  | 19 |  | ab | 10 |  |  | ${ }_{11}$ |  |

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|  | a) | $\mathrm{a}^{\prime} 0^{\circ}$ | $\mathrm{a}^{\prime} 0$ | a | ac ${ }^{1}$ | 0.0 | $0^{\prime} 0$ | $0^{\prime} 0$ | 00 | 00 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $b c$ |  | $00$ | 01 | 11 | 10 | C? | $00$ | 01 | 11 | 10 |
| $0^{\prime} C^{\prime}$ | 00 | 0 | 1 | 9 | 8 | $c^{\prime} 0^{\prime} 00$ | 0 | 4 | 5 | 1 |
| $0^{\circ} \mathrm{C}$ | $01$ | 2 | 3 | 11 | 10 | $c^{\prime} a$ | - 8 | $12$ | 13 | 9 |
| $0 ¢$ | $11$ | 6 | 7 | 15 | 14 |  | 10 | 14 | 15 | 11 |
| $b c^{\prime}$ | $10$ | 4 | 5 | 13 | 12 | $\text { ca } 10$ | 2 | 6 | 7 | 3 |

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|  | ab | $\mathbf{a}^{\prime} \mathbf{b}^{\prime}$ | $\mathbf{a}^{\prime} \mathbf{b}$ | ab | $\mathbf{a b} \mathbf{b}^{\prime}$ |
| :---: | :---: | ---: | ---: | ---: | ---: |
| c |  | 00 | 01 | 11 | 10 |
| c' $^{\prime}$ | 0 | 0 | 2 | 6 | 4 |
| c | 1 | 1 | 3 | 7 | 5 |


|  | $b c$ | $b^{\prime} c^{\prime}$ | $b^{\prime} c$ | $b c$ | $b c^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{a}$ |  | 00 | 01 | 11 | 10 |
| $a^{\prime}$ | 0 |  |  | 3 | 2 |
| $\mathbf{a}$ | 1 | 4 | 5 | 7 | ${ }_{6}$ |


|  | $a$ | $a^{\prime}$ | $a$ |
| :---: | :---: | :---: | :---: |
| $b c$ |  | 0 | 1 |
| $b^{\prime} c^{\prime}$ | 00 | 0 | 4 |
| $b^{\prime} c$ | 01 | 1 | 5 |
| $b c$ | 11 | 3 | 7 |
| $a c^{\prime}$ | 10 | 2 | ${ }_{6}$ |

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## 5-Variable K-Map


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## 6-Variable K-Map



## 7-Variable K-Map

| $\mathbf{c}^{\prime} \mathbf{0}^{\prime} c^{\prime}$ |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | $c d$ | $c^{\prime} d^{\prime}$ | $c^{\prime} d$ | $c d$ | $c d^{\prime}$ |
| ef |  | 00 | 01 | 11 | 10 |
| $e^{\prime} f^{\prime}$ | 00 |  |  |  |  |
| e'f | 01 |  | 0 | -1.5 |  |
| ef | 11 |  |  |  |  |
| eff | 10 |  |  |  |  |


| $a^{\prime \prime} 0^{\prime} \mathrm{C}$ |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | cd | $c^{\prime} \mathrm{d}^{\prime}$ | c'd | cd | cd' |
| ef |  | 00 | 01 | 11 | 10 |
| e'f' | 00 |  |  |  |  |
| e'f | 01 | 16-31 |  |  |  |
| ef | 11 |  |  |  |  |
| ef' | 10 |  |  |  |  |


abe

|  | $c d$ | $c^{\prime} d^{\prime}$ | $c^{\prime} d$ | $c d$ | $c d^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ef |  | 00 | 01 | 11 | 10 |
| $e^{\prime} f^{\prime}$ | 00 |  |  |  |  |
| $e^{\prime} f$ | 01 |  | 64 | -7 |  |
| ef | 11 |  |  |  |  |
| eff | 10 |  |  |  |  |


abc

|  | $c d$ | $c^{\prime} d^{\prime}$ | $c^{\prime} d$ | $c d$ | $c d^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ef |  | 00 | 01 | 11 | 10 |
| $e^{\prime} f^{\prime}$ | 00 |  |  |  |  |
| e'f $^{\prime}$ | 01 | 1 | 2 | 12 |  |
| ef | 11 |  | 2 | 2 |  |
| ef | 10 |  |  |  |  |

$a^{\prime} b^{\prime} c^{\prime} d^{\prime}$

$a^{\prime} b c^{\prime} d^{\prime}$

|  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| ${ }_{\text {efp }}$ |  |  |  |  |
|  |  |  |  |  |
|  |  |  |  |  |

$a b^{\prime} c^{\prime} d^{\prime}$

|  | $c d$ | $c^{\prime} d^{\prime}$ | $c^{\prime} d$ | $c d$ | $c d^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ef |  | 00 | 01 | 11 | 10 |
| $e^{\prime} f^{\prime}$ | 00 |  |  |  |  |
| e'f | 01 |  | 0 | 1 | 3 |
| ef | 11 |  |  |  |  |
| ef' | 10 |  |  |  |  |

$a b c^{\prime} d^{\prime}$

 off oi 192-207
$a^{\prime} b^{\prime} c^{\prime} d$

|  | cd | $\mathrm{c}^{\prime} \mathrm{d}^{\prime}$ | c'd | cd | cd' |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ef |  | 00 | 01 | 11 | 10 |
| e'f' | 00 | $16-31$ |  |  |  |
| e'f | 01 |  |  |  |  |
| ef | 11 |  |  |  |  |
| ef' | 10 |  |  |  |  |

$a^{\prime} b c^{\prime} d$

|  |
| :---: |
|  |
| eff ol 80-95 |

$a b^{\prime} c^{\prime} d$

|  | $c d$ | $c^{\prime} d^{\prime}$ | $c^{\prime} d$ | $c d$ | $c d^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ef |  | 00 | 01 | 11 | 10 |
| $e^{\prime} f^{\prime}$ | 00 |  |  |  |  |
| $e^{\prime} f$ | 01 |  |  |  |  |
| ef | 11 |  |  |  |  |
|  |  |  |  |  |  |

abc'd

|  |
| :---: |
|  |  |
|  |  |

$a^{\prime} b^{\prime} c d^{\prime}$

$a^{\prime} b c d^{\prime}$

$a^{\prime} b^{\prime} c d$

ab'cd

|  | $c d$ | $c^{\prime} d^{\prime}$ | $c^{\prime} d$ | $c d$ | $c d^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ef |  | 00 | 01 | 11 | 10 |
| $e^{\prime} f^{\prime}$ | 00 |  |  |  |  |
| $e^{\prime} f$ | 01 |  | 6 | 1 | 1 |
| ef | 11 | 1 |  |  | 1 |
| eff | 10 |  |  |  |  |

abcd

e'p 00
eff
of
11 240-255

## K-Map for POS

|  | a+b | $a+b$ | $a+b^{\prime}$ | $a^{\prime}+b^{\prime}$ | $a^{\prime}+b$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| c+d |  | 0+0 | 0+1 | 1+1 | $1+0$ |
| c+d | 0+0 |  |  | , |  |
| c+d' | 0+1 |  |  | 13 |  |
| $c^{\prime}+\mathrm{d}^{\prime}$ | $1+1$ |  |  | 15 |  |
| $c+d$ | $1+0$ |  |  |  |  |

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- Minimal function: - A function is said to be minimal if it is representing the function expression, if it is using minimal no of literals to represent the function.
- Rules of grouping: -

1. Every minterm must be covered
2. Group must have contiguous Cells(circular)
3. Group must be in horizontal or vertical fashion(circular)
4. Number of cells in a group must be in power of 2
5. Will Try to make largest group possible, so that number of literals in the expression can be reduced
6. Can also take don't care if it helps in creating the larger groups, other wise don't care.
7. Will consider new implicant if it is covering some new minterm.

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## Don't care condition

- Don't care cases are those cases which can never occur logically in that function.
- It is not essential to cover don't care conditions, but if don't care are helping to generate bigger prime implicants, then we can use them.


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|  | $a b$ | $a^{\prime} b^{\prime}$ | $a^{\prime} b$ | $a b$ | $a b^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $c d$ |  | 00 | 01 | 11 | 10 |
| $c^{\prime} d^{\prime}$ | 00 |  |  | 1 |  |
| $c^{\prime} d$ | 01 | $D$ | $D$ | 1 |  |
| $c c^{\prime}$ | 11 |  | 1 | 1 |  |
| $c^{\prime}$ | 10 |  | 1 | 1 |  |

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First try to cover those minterms which do not have an option.

|  | $a b$ | $a^{\prime} b^{\prime}$ | $a^{\prime} b$ | $a b$ | $a b^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $c d$ |  | 00 | 01 | 11 | 10 |
| $c^{\prime} d^{\prime}$ | 00 |  | 1 |  |  |
| $c^{\prime} d$ | 01 |  | 1 | 1 | 1 |
| $c d$ | 11 | 1 | 1 | 1 |  |
| $c^{\prime} d^{\prime}$ | 10 |  |  | 1 |  |

For some functions more than one minimal Boolean expression are possible

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$\mathbf{Q} f(a, b, c)=\sum_{m}\{1,2,3,4,5\}$

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|  | $a b$ | $a^{\prime} b^{\prime}$ | $a^{\prime} b$ | $a b$ | $a b^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| cd |  | 00 | 01 | 11 | 10 |
| $c^{\prime} d^{\prime}$ | 00 | 1 |  |  |  |
| c'd | 01 | 1 | 1 |  |  |
| cd | 11 |  | 1 | 1 |  |
|  |  |  |  |  |  |
| cd' | 10 |  |  | 1 |  |

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|  | ab | $a^{\prime} b^{\prime}$ | $a^{\prime} b$ | ab | ab' |
| :---: | :---: | :---: | :---: | :---: | :---: |
| cd |  | 00 | 01 | 11 | 10 |
| $c^{\prime} d^{\prime}$ | 00 | 1 | $D$ | D | 1 |
| $c^{\prime} d$ | 01 | D |  |  |  |
| cd | 11 |  |  |  |  |
| cd' | 10 | 1 |  |  |  |
|  |  |  |  |  |  |

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|  | $a b$ | $a^{\prime} b^{\prime}$ | $a^{\prime} b$ | $a b$ | $a b^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $c d$ |  | 00 | 01 | 11 | 10 |
| $c^{\prime} d^{\prime}$ | 00 | $D$ | 1 |  | 1 |
| $c^{\prime} d$ | 01 |  | 1 | $D$ |  |
| $c d$ | 11 | 1 | $D$ | $D$ |  |
| $c^{\prime} d^{\prime}$ | 10 | $D$ |  |  |  |

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K-Map for POS

|  | $a+b$ | $a+b$ | $a+b^{\prime}$ | $a^{\prime}+b^{\prime}$ | $a^{\prime}+b$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $c+d^{\prime}$ |  | $0+0$ | $0+1$ | $1+1$ | $1+0$ |
| $c+d^{\prime}$ | $0+0$ |  |  | $D$ | 1 |
| $c+d^{\prime}$ | $0+1$ | 1 | $D$ | 1 |  |
| $c^{\prime}+d^{\prime}$ | $1+1$ | 1 | $D$ | $D$ |  |
| $c^{\prime}+d^{\prime}$ | $1+0$ | 1 | $D$ | $D$ |  |

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Using Quine Mc-Clusky Method(Tabular Method)


| Group | Minterm | Binary Representation | Group | Minterm | Binary Representation |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | $\begin{aligned} & 0,1 \\ & 0,8 \end{aligned}$ | $\begin{array}{r} 000 \\ 000 \\ \hline \end{array}$ | 0 | $\begin{aligned} & 0,1,8,9 \\ & 0,8,1,9 \end{aligned}$ | $\begin{aligned} & 00 \\ & -00- \end{aligned}$ |
| 1 | 1,3 | $00 \_1$ |  |  |  |
|  | 1,9 | _ 001 | 1 | $\begin{aligned} & 1,3,9,11 \\ & 1,9,3,11 \end{aligned}$ | - 0 - 1 |
|  | 8,9 | 100 |  |  | - 0 _ 1 |
| 2 | 3,7 | $0 \_11$ |  |  |  |
|  | 3,11 | _011 |  |  |  |
|  | 9,11 | $10-1$ | 2 |  | - - 11 |
| 3 | 7,15 | - -111 |  | $3,11,7,15$ | _ _ 11 |
|  | 11,15 | $1_{\text {_ }} 11$ |  |  |  |

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| Group | Minterm | Binary <br> Representation |  |
| :---: | :---: | :---: | :---: |
| 0 | $0,1,8,9$ <br> $0,8,1,9$ | $-00_{-}$ <br> $-00_{-}$ | $b^{\prime} c^{\prime}$ |
| 1 | $1,3,9,11$ <br> $1,9,3,11$ | $-0-1$ <br> -0 | b'd $^{\prime} 1$ |

$Y(a, b, c, d)=b^{\prime} c^{\prime}+c d$

| PI | Minterm | 0 | 1 | 3 | 7 | 8 | 9 | 11 | 15 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{b}^{\prime} \mathrm{c}^{\prime}$ | 0,1,8,9 | x | X |  |  | X | X |  |  |
| b'd | 1,3,9,11 |  | x | x |  |  | X | x |  |
| cd | 3,7,11,15 |  |  | X | x |  |  | x | X |
| www.knowledgegate.in |  |  |  |  |  |  |  |  |  |



## Combinational Circuits



$$
0 / p=f(i / p)
$$

## LET'S GET STARTED!


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## Combinational Circuits

- When logic gates are connected together to produce a specified out put on certain specified combinations of input variables, with no memory involved, then the resulting circuit is called a combinational circuit.
- Output depends only on present input. $O / p=f(i / p)$, combinational circuit performs an operation that can be specified logically by a set of Boolean function.
- A combinational circuit may have $n$-binary inputs and m-binary outputs.

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- Application, Practical computer circuits normally contain combinational and sequentiallogic. For e.g. the part of ALU, that does mathematical calculations is constructed using combinational logic.

- Other circuits such as half adders, full adders, half subtractors, full subtractors, multiplexers, demultiplexers, encoders and decoders are also made by using combinational logic.



## wow

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Design procedure:

1. Analyse the given problem and identify the number of $i / p$ and $o / p$ variables.
2. Write the truth table based on the specification of the problem.
3. Convert the truth table in minimized Boolean expression using k-map.
4. Draw the logic circuit for the above obtained output expression.

Q Design a digital system for a car manufacturing company, where we want to design awarning signal for a car, there are three inputs, lights of the car(L), day or night(D), ignition (on/off).? Solution

1) Understand the problem- here we will Understand the definition of the problem Design the truth table -

| Light | Day | Engine | Warning |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 |  |
| 0 | 0 | 1 |  |
| 0 | 1 | 0 |  |
| 0 | 1 | 1 |  |
| 1 | 0 | 0 |  |
| 1 | 0 | 1 |  |
| 1 | 1 | 0 |  |
| 1 | 1 | 1 |  |

2) Write the Boolean expression

- $W(L, D, E)=\Sigma_{m}(1,4,6,7)$
- $W(L, D, E)=\Pi_{M}(0,2,3,5)$

| 0 | 0 | 0 | 0 |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 0 | 1 |
| 1 | 1 | 1 | 1 |

3) Minimize Boolean expression $\mathrm{W}=\mathrm{a}^{\prime} \mathrm{b}^{\prime} \mathrm{c}+\mathrm{ab} \mathrm{c}^{\prime}+\mathrm{abc}+\mathrm{abc}$

|  | ab | $a^{\prime} b^{\prime}$ | $a^{\prime} b$ | $a b$ | $a b^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{c}$ |  | 00 | 01 | 11 | 10 |
| $\mathrm{c}^{\prime}$ | 0 | 0 | 2 | 6 | 4 |
| c | 1 | 1 | 3 | 7 | 5 |

W =

| Light | Day | Engine | Warning |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 0 | 1 |
| 1 | 1 | 1 | 1 |

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| 4) Implement the expression using | Light | Day | Engine | Warning |
| :---: | :---: | :---: | :---: | :---: |
| logic gates, draw the | 0 | 0 | 0 | O |
| implementation using logic gates | 0 | 0 | 1. | 1 |
| $\mathrm{W}=\mathrm{ac}$ + $\mathrm{ab}+\mathrm{a}^{\prime} \mathrm{b}^{\prime} \mathrm{c}$ | 0 | 1 | - 0 | 0 |
|  | 0 | 15 | 1 | 0 |
|  | 1. | 0 | 0 | 1 |
|  | 1 | 0 | 1 | 0 |
|  | 1 | 1 | 0 | 1 |
|  | 1 | 1 | 1 | 1 |

## Adder

- An adder is a digital combinational circuit that performs addition of numbers. Are used in the arithmetic logic units or ALU.
- They are also utilized in other parts of the processor, where they are used to calculate addresses, table indices, increment and decrement operators, and similar operations.
- Although adders can be constructed for many number representations, such as-binary-coded decimal or excess3 , the most common adders operate on binary numbers.
- Can also be used to perform substruction In cases of 2's representation.
$\leqslant 4$


## Half adder

- The simplest form of addition is addition of two binary digits, consists of four possible elementary operations


## 1



- The first three operations produce a sum of two digits, but when both augend and addend bits are equal to 1 , the binary sum consists of two digits. The higher significant bit of this result is called a carry.
- It is a combinational circuit, which perform the arithmetic addition of two one-bit binary numbers is referred to as an half-adder.
- So, in half adder inputs are adds two single binary bits $A$ and $B$, and two outputs, sum $(S)$ and carry (C).


| INPUTS |  | OUTPUTS |  |
| :---: | :---: | :---: | :---: |
| A | B | Carry | Sum |
| 0 | 0 |  |  |
| 0 | 1 |  |  |
| 1 | 0 |  |  |
| 1 | 1 |  |  |

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- Cost of implementation a half adder is one EX-OR gate and one AND gate.
- A half adder has only two inputs and there is no provision to add a carry coming from the lower order bits when multi bit number addition is performed. For this reason, we have désigned a full adder.
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## Half subtractor



| Inputs |  | Outputs |  |
| :---: | :---: | :---: | :---: |
| A | B | Diff | Borrow |
| 0 | 0 | 0 | 0 |
| 0 | 1 | 1 | 1 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 0 | 0 |



## Full adder

1. A full adder is a combinational logic circuit that performs the arithmetic sum of three input bits.
2. Where $A_{n}, B_{n}$ are the $n^{\text {th }}$ order bits of the number $A$ and $B$ respectively and $C_{n}$ is the carry generated from the addition of $(\mathrm{n}-1)^{\text {th }}$ order bits.
3. It consists of three input bits, denoted by A (First operand), B (Second operand), $\mathrm{C}_{\text {in }}$ (Represents carry from the previous lower significant position).
(C43

- Two output bits are same as of half adder, which is Sum and Carry out.
- When the augend and addend number contain more significant digits, the carry obtained from the addition of two bits is added to the next higher order pair of significant bits.


| INPUTS |  |  | OUTPUTS |  |
| :---: | :---: | :---: | :---: | :---: |
| $A$ | $B$ | $C_{\text {in }}$ | $C_{\text {out }}$ | Sum |
| 0 | 0 | 0 |  |  |
| 0 | 0 | 1 |  |  |
| 0 | 1 | 0 |  |  |
| 0 | 1 | 1 |  |  |
| 1 | 0 | 0 |  |  |
| 1 | 0 | 1 |  |  |
| 1 | 1 | 0 |  |  |
| 1 | 1 | 1 |  |  |

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|  | ab | $a^{\prime} b^{\prime}$ | $a^{\prime} b$ | ab | ab |
| :---: | :---: | :---: | :---: | :---: | ---: |
| $\mathrm{c}_{\text {in }}$ |  | 00 | 01 | 11 | 10 |
| $\mathrm{c}_{\text {in }}^{\prime}$ | 0 | 0 | 2 | 6 | 4 |
| $\mathrm{c}_{\text {in }}$ | 1 | 1 | 1 | 3 | 7 |


|  | $a b$ | $a^{\prime} b^{\prime}$ | $a^{\prime} b$ | $a b$ | $a b^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $c_{\text {in }}$ |  | 00 | 01 | 11 | 10 |
| $c_{\text {in }}^{\prime}$ | 0 | 0 | 2 | 6 |  |
| $c_{\text {in }}$ | 1 |  | 3 | 7 | 5 |


| INPUTS |  |  | OUTPUTS |  |
| :---: | :---: | :---: | :---: | :---: |
| $A$ | $B$ | $C_{\text {in }}$ | Sum | $C_{\text {out }}$ |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 | 0 |
| 0 | 1 | 0 | 1 | 0 |
| 0 | 1 | 1 | 0 | 1 |
| 1 | 0 | 0 | 1 | 0 |
| 1 | 0 | 1 | 0 | 1 |
| 1 | 1 | 0 | 0 | 1 |
| 1 | 1 | 1 |  |  |
| 1 |  |  |  |  |



Full subtractor


| Inputs |  |  | Outputs |  |
| :---: | :---: | :---: | :---: | :---: |
| A | B | Bofrow $_{\text {in }}$ | Diff | Borrow |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 | 1 |
|  | 1 | 0 | 1 | 1 |
| 0 | 1 | 1 | 0 | 1 |
| 1 | 0 | 0 | 1 | 0 |
| 1 | 0 | 1 | 0 | 0 |
| 1 | 1 | 0 | 0 | 0 |
| 1 | 1 | 1 | 1 | 1 |

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## Four-bit parallel binary adder / Ripple adder

- As we know that full adder is capable of adding two 1 bit number and 1 previous carry, but in order to add binary numbers with more than one bits, additional full adders must be employed. For e.g. a four bit binary adder can be constructed using four full adders.
- Theses four full adders are connected in cascade, carry output of each adder is connected to the carry input of the next higher-order adder. So a $n$-bit parallel adder is constructed using " $n$ ' number of full adders.



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- There are some scope of improvement in parallel binary adder / Ripple adder is it is very slow

Carry propagation delay
Look ahead Carry Generator


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## Look ahead carry adder

- The carry propagation time is an important attribute of the adder because it limits the speed with which two numbers are added. The solution to delay is to increase the complexity of the equipment in such a way that the carry delay time is reduced.
- To solve this problem most widely used technique employs the principle of 'look ahead carry'. This method utilizes logic gates to look at the lower order bits of the augend and addend to see if a higher order carry is to be generated. It uses two functions carry generate $G_{i}$ and carry propagate $P_{i}$

$A\left(A_{3} A_{2} A_{1} A_{0}\right)$ $B\left(B_{3} B_{2} B_{1} B_{0}\right)$
- $G_{i}$ is called a carry generate, and it produces a carry of 1 when both $A_{i}$ and $B_{i}$ are 1 , regardless of the input carry $C_{i}$
- $P_{i}$ is called a carry propagate, because it determines whether a carry into stage i will propagate into stage $\mathrm{i}+1$.
- We now write the Boolean functions for the carry outputs of each stage and substitute the value of each $\mathrm{C}_{i}$ from the previous equations:
- $P_{i}=A_{i} \oplus B_{i}$
$\mathrm{G}_{\mathrm{i}}=\mathrm{A}_{\mathrm{i}} . \mathrm{B}_{\mathrm{i}}$
$S_{i}=P_{i} \oplus C_{i}$
$C_{i+1}=G_{i}+P_{i} C_{i}$
- $\mathrm{C}_{0}=0$
- $\mathrm{C}_{1}=\mathrm{G}_{0}+\mathrm{P}_{0} \mathrm{C}_{0}$

- $\mathrm{C}_{2}=\mathrm{G}_{1}+\mathrm{P}_{1} \mathrm{G}_{0}+\mathrm{P}_{1} \mathrm{P}_{0} \mathrm{C}_{0}$
- $C_{3}=G_{2}+P_{2} G_{1}+P_{2} P_{1} G_{0}+P_{2} P_{1} P_{0} C_{0}$
$C_{4}=G_{3}+G_{2} \cdot P_{3}+G_{1} \cdot P_{2} \cdot P_{3}+G_{0} \cdot P_{1} \cdot P_{2} \cdot P_{3}+C_{0} \cdot P_{0} \cdot P_{1} \cdot P_{2} \cdot P_{3}$


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- Since the Boolean function for each output carry is expressed in sum-of-products form. Each function can be implemented with one level of AND gates followed by an OR gate (or by a twolevel NAND).
- $\mathrm{C}_{0}=0$
- $\mathrm{C}_{1}=\mathrm{G}_{0}+\mathrm{P}_{0} \mathrm{C}_{0}$
- $\mathrm{C}_{2}=\mathrm{G}_{1}+\mathrm{P}_{1} \mathrm{G}_{0}+\mathrm{P}_{1} \mathrm{P}_{0} \mathrm{C}_{0}$
- $\mathrm{C}_{3}=\mathrm{G}_{2}+\mathrm{P}_{2} \mathrm{G}_{1}+\mathrm{P}_{2} \mathrm{P}_{1} \mathrm{G}_{0}+\mathrm{P}_{2} \mathrm{P}_{1} \mathrm{P}_{0} \mathrm{C}_{0}$
- $\mathrm{C}_{4}=\mathrm{G}_{3}+\mathrm{G}_{2} \cdot \mathrm{P}_{3}+\mathrm{G}_{1} \cdot \mathrm{P}_{2} \cdot \mathrm{P}_{3}+\mathrm{G}_{0} \cdot \mathrm{P}_{1} \cdot \mathrm{P}_{2} \cdot \mathrm{P}_{3}+\mathrm{C}_{0} \cdot \mathrm{P}_{0} \cdot \mathrm{P}_{1} \cdot \mathrm{P}_{2} \cdot P_{3}$


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- All output carries are generated after a delay through two levels of gates. Thus, outputs $\mathrm{S}_{1}$ through $\mathrm{S}_{3}$ have equal propagation delay times.



## Four-bit ripple adder/subtractor

- The subtraction $A-B$ can be done by taking the 2 's complement of $B$ and adding it to $A$.
- $A+(-B)$



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- The circuit for subtracting A - B consists of an adder with inverters placed between each data input B and the corresponding input of the full adder. The mode input $M$ controls the operation. When $M=0$, the circuit is an adder, and when $\mathrm{M}=1$, the circuit becomes a subtractor.
- When $M=0$, we have $B \oplus 0=B$. The full adders receive the value of $B$, the input carr $y$ is 0 , and the circuit performs A plus $B$.
- When $M=1$, we have $B \oplus 1=B^{\prime}$ and $C_{0}=1$. The $B$ inputs are all complemented and a 1 is added through the input carry. The circuit performs the operation A plus the 2's complement of B.


|  |  |  |  |  |  |  | $A>B$ | $a_{1} a_{0}$ | $\mathrm{a}^{\prime} \mathrm{a}^{\prime}$ | $\mathrm{a}_{1} \mathrm{a}_{0}$ | $\mathrm{a}_{4} \mathrm{a}_{0}$ | $\mathrm{a}_{1} \mathrm{a}^{\prime}{ }_{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{A}_{1}$ | $A_{0}$ | $B_{1}$ | $B_{0}$ | $A>B$ | $A<B$ | $A=B$ | $\mathrm{b}_{1} \mathrm{~b}_{0}$ |  | 00 | 01 | 11 | 10 |
| 0 | 0 | 0 | 0 | 0 | 0 | 1 | $\mathrm{b}^{\prime} \mathrm{b}^{\prime} \mathrm{b}_{0}$ | 00 |  | 1 | 1 | 11 |
| 0 | 0 | 0 | 1 | 0 | 1 | 0 | $\mathrm{b}^{\prime} \mathrm{b}_{0} \mathrm{~b}_{0}$ | 01 |  |  | 1 | 1 |
| 0 | 0 | 1 | 0 | 0 | 1 | 0 | $\mathrm{B}_{1} \mathrm{~b}_{0}$ | 11 |  |  |  |  |
| 0 | 0 | 1 | 1 | 0 | 1 | 0 | $\mathrm{B}_{1} \mathrm{~b}^{\prime}{ }_{0}$ | 10 |  |  | 1 |  |
| 0 | 1 | 0 | 0 | 1 | 0 | 0 | A $=\mathrm{B}$ | $\mathrm{a}_{1} \mathrm{a}_{0}$ | $\mathrm{a}^{\prime} \mathrm{a}^{\prime}$ | $\mathrm{a}_{1} \mathrm{a}_{0}$ | $\mathrm{a}_{1} \mathrm{a}_{0}$ | $\mathrm{a}_{1} \mathrm{a}^{\prime}$ |
| 0 | 1 | 0 | 1 | 0 | 0 | 1 | $b_{1} b_{0}$ |  | 00 | 01 | 11 | 10 |
| 0 | 1 | 1 | 0 | 0 | 1 | 0 | $\mathrm{b}^{\prime}{ }_{1} \mathrm{~b}^{\prime}{ }_{0}$ | 00 | 1 |  |  |  |
| 0 | 1 | 1 | 1 | 0 | 1 | 0 | $\mathrm{b}^{\prime} \mathrm{b}^{\text {b }}$ 0 | 01 |  | 1 |  |  |
| 1 | 0 | 0 | 0 | 1 | 0 | 0 | $\mathrm{B}_{1} \mathrm{~b}_{0}$ | 11 |  |  | 1 |  |
| 1 | 0 | 0 | 1 | 1 | 0 | 0 | $\mathrm{B}_{1} \mathrm{~b}^{\prime}{ }_{0}$ | 10 |  |  |  | 1 |
| 1 | 0 | 1 | 0 | 0 | 0 | 1 | A<B | $\mathrm{a}_{1} \mathrm{a}_{0}$ | $a^{\prime} a^{\prime}$ | $\mathrm{a}^{\prime} \mathrm{al}^{1}$ | $a_{1} a_{0}$ | $\mathrm{a}_{1} \mathrm{a}_{0}$ |
| 17 | 0 | 1 | 1 | 0 | 1 | 0 | $\mathrm{b}_{1} \mathrm{~b}_{0}$ |  | 00 | 01 | 11 | 10 |
| 1 | 1. | 0 | 0 | 1 | 0 | 0 | $\mathrm{b}^{\prime} \mathrm{l}^{\prime} \mathrm{b}_{0}$ | 00 |  |  |  |  |
| $1 \checkmark$ | 1 | 0 | 1 | 1 | 0 | 0 | $\mathrm{b}^{\prime}, \mathrm{b}_{0}$ | 01 | 1 |  |  |  |
| 1 | 1 | 1 | 0 |  |  |  | $\mathrm{B}_{1} \mathrm{~b}_{0}$ | 11 | 1 | 1 |  | 1 |
| 1 | 1 | 1 | 1 | NOW | . | $W_{1}$ | $\mathrm{B}_{1} \mathrm{~b}^{\prime}{ }_{0}$ | 10 | 1 | 1 |  |  |


| $A>B$ | $a_{1} a_{0}$ | $a_{1}^{\prime} a^{\prime}$ | $a_{1} a_{0}$ | $a_{1} a_{0}$ | $\mathrm{a}_{1} \mathrm{a}_{0}^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{b}_{1} \mathrm{~b}_{0}$ |  | 00 | 01 | 11 | 10 |
| $\mathrm{b}_{1}{ }_{1} \mathrm{~b}^{\prime}$ | 00 |  | 1 | 1 | 1 |
| $\mathrm{b}_{1}^{\prime} \mathrm{b}_{0}$ | 01 |  |  | 1 | 1 |
| $\mathrm{B}_{1} \mathrm{~b}_{0}$ | 11 |  |  |  |  |
| $\mathrm{B}_{1} \mathrm{~b}^{\prime}{ }_{0}$ | 10 |  |  | 1 |  |


| $A=B$ | $a_{1} a_{0}$ | $a_{1}^{\prime} a_{0}^{\prime}$ | $a^{\prime}{ }_{1} a_{0}$ | $a_{1} a_{0}$ | $a_{1} a_{0}^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $b_{1} b_{0}$ |  | 00 | 01 | 11 | 10 |
| $b_{1}^{\prime} b_{0}^{\prime}$ | 00 | 1 |  |  |  |
| $b_{1}^{\prime} b_{0}$ | 01 |  | 1 |  |  |
| $B_{1} b_{0}$ | 11 |  |  | 1 |  |
| $B_{1} b_{0}^{\prime}$ | 10 |  |  |  | 1 |

$A>B=a_{0} b_{1} b_{0}+a_{1} a_{1} b_{1}+a_{1} b_{1}$
$A=B=\left(a_{1} \odot b_{1}\right)\left(a_{0} \odot b_{0}\right)$

| $A<B$ | $a_{1} a_{0}$ | $a^{\prime} a_{0}^{\prime}$ | $a_{1}^{\prime} a_{1} 1_{0}$ | $a_{1} a_{0}$ | $a_{1} a_{0}^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $b_{1} b_{0}$ |  | 00 | 01 | 11 | 10 |
| $b_{1}^{\prime} b_{0}^{\prime}$ | 00 |  |  |  |  |
| $b_{1}^{\prime} b_{0}$ | 01 | 1 |  |  |  |
| $B_{1} b_{0}$ | 11 | 1 | 1 |  | 1 |
| $B_{1} b_{0}^{\prime}$ | 10 | 1 | 1 |  |  |

## 




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## Multiplexer (Selector)

- Multiplexer are special and one of the most widely used combinational circuits.
- Main requirement is out of many inputs we have to select one for e.g. telephone or train leaving the station. So multiplexer do not perform any logical operation or comparison, it just acts as a switch or relay.



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1. A multiplexer is a combinational circuit that selects binary information from one of many input lines and directs it to a single output line.
2. A multiplexer is also called a data selector, since it selects one of many inputs and steers the binary information to the output line.
3. The selection of a particular input line is controlled by a set of selection lines.
4. There are $2^{n}$ input lines and $n$ selection lines whose bit combinations determine which input is to be selected.

## Note:

- Can never have two i/p connected to out at any time


## Application

- Parallel data to serial data conversion
- Used as data selector, as the output of a multiplexer is directed from one various inputs
- Used in implementation of Boolean functions
- Used in communication systems, Computer Memory, Telephone Network, Transmission from the Computer System of a Satellite


## A two-to-one-line multiplexer

- When $\mathrm{S}=0$, the upper AND gate is enabled and $\mathrm{I}_{0}$ has a path to the output.
- When $S=1$, the lower AND gate is enabled and $I_{1}$ has a path to the output.


| Input | Output |
| :---: | :---: |
| S | Y |
| 0 | $\mathrm{I}_{0}$ |
| 1 | $I_{1}$ |



Characteristic equation: $\mathrm{Y}=\mathrm{S}_{0}^{\prime} \mathrm{I}_{0}+\mathrm{S}_{0} \mathrm{I}_{1}$
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## Case study of 4 to 1

- Each of the four inputs, $I_{0}$ through $I_{3}$, is applied to one input of an AND gate.
- Selection lines $\mathrm{S}_{1}$ and $\mathrm{S}_{0}$ are decoded to select a particular AND gate.
- The outputs of the AND gates are applied to a single OR gate that provides the one-line output.


| Input |  | Output |
| :---: | :---: | :---: |
| $\mathrm{S}_{1}$ | $\mathrm{~S}_{0}$ | Y |
| 0 | 0 | $\mathrm{I}_{0}$ |
| 0 | 1 | $\mathrm{I}_{1}$ |
| 1 | 0 | $\mathrm{I}_{2}$ |
| 1 | 1 | $\mathrm{I}_{3}$ |



Characteristic equation
$Y=E\left[\left(S_{1}^{\prime} S_{0}^{\prime} I_{0}\right)+\left(S_{1}^{\prime} S_{0} I_{1}\right)+\left(S_{1} S_{0}^{\prime} I_{2}\right)+\left(S_{1} S_{0} I_{3}\right)\right]$

Q Implement the function $F(a, b, c)=a^{\prime} b^{\prime} c+a^{\prime} b c^{\prime}+a b^{\prime} c^{\prime}+a b c$. With $b$ and c variable on select line?


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$\mathbf{Q}$ Implement the function $\mathrm{F}(\mathrm{a}, \mathrm{b}, \mathrm{c})=\mathrm{a}^{\prime} \mathrm{b}^{\prime} \mathrm{c}+\mathrm{a}^{\prime} \mathrm{bc} c^{\prime}+\mathrm{ab} \mathrm{b}^{\prime} \mathrm{c}^{\prime}+\mathrm{abc}$. With one, two and three variable on the select line?

## Demultiplexer

1. A demultiplexer (or DeMux) is a device that takes a single input line and routes it to one of several digital output lines.
2. A demultiplexer is also called a data distributor.
3. It is conceptually same as Mux just with reverse logic.

- A demultiplexer of $2^{n}$ outputs has $n$ select lines, which are used to select which output line to send the input.
- A DeMux is a combinational circuit, which is used in data communication
- Serial to parallel conversation
- Demultiplexers are mainly used in Boolean function generators and decoder circuits.


## 1 to 2 Demultiplexer



| $\mathrm{S}_{0}$ | $\mathrm{O}_{1}$ | $\mathrm{O}_{0}$ |
| :---: | :---: | :---: |
| 0 | 0 | 1 |
| 1 | 1 | 0 |


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1 to 4 Demultiplexer

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## Decoder

- A decoder is a combinational circuit that decodes binary information from $n$ input lines to a maximum of $2^{n}$ unique output lines.
- The decoders are called $n$-to- $m$-line decoders, where $m \leq 2^{n}$.
- Their purpose is to generate the $2^{n}$ (or fewer) minterms of $n$ input variables. Each combination of inputs will assert a unique output.
- If the n -bit coded information has unused combinations, the decoder may have fewer than $2^{n}$ outputs.


Decoder

- Decoders are also combinational circuits logically we can say a DeMux can be converted into a decoder by setting input line as enable line and selection line as input lines.


Address Bus - $\mathrm{A}_{3}$ to $\mathrm{A}_{10}$



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## 1-to-2 Decoder

\section*{$I-$| Decadn |
| :---: |
| $E$ | $0^{-0,}$}


| I | $\mathrm{O}_{1}$ | $\mathrm{O}_{0}$ |
| :---: | :---: | :---: |
| 0 | 0 | I |
| 1 | I | 0 |


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## 2-to-4 Decoder


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## Q Implementation of a full adder with a decoder

 From the truth table of the full adder, we obtain the functions for the combinational circuit in sum-of-minterms form:$S(x, y, z)=\sum(1,2,4,7)$ $C(x, y, z)=\sum(3,5,6,7)$


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## Combinational Logic Implementation

- Any combinational circuit with $n$ inputs and $m$ outputs can be implemented with an $n$-to-2 line decoder and OR gates.
- The procedure for implementing a combinational circuit by means of a decoder and OR gates requires that the Boolean function for the circuit be expressed as a sum of minterms.
- A decoder is then chosen that generates all the minterms of the input variables. The inputs to each OR gate are selected from the decoder outputs according to the list of minterms of each function.
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## Q A full Subtractor using Decoder?


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## Encoder

- An encoder is a combinational circuit that encode binary information form one of a $2^{\mathrm{N}}$ input lines and encode it into N output lines, which represent $N$ bit code for the input.
- For simple encoders, it is assumed that only one input line is active at a time.
- Encoder performs the inverse operation of a decoder.


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## 2-to-1 Encoder



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## 4-to-2 Encoder



| $\mathrm{I}_{3}$ | $\mathrm{I}_{2}$ | $\mathrm{I}_{1}$ | $\mathrm{I}_{0}$ | $\mathrm{O}_{1}$ | $\mathrm{O}_{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 1 | 0 | 0 |
| 0 | 0 | 1 | 0 | 0 | 1 |
| 0 | 1 | 0 | 0 | 1 | 0 |
| 1 | 0 | 0 | 0 | 1 | 1 |$\quad \mathrm{O}_{0}=\mathrm{I}_{0}{ }^{\prime} \mathrm{I}_{1}{ }^{\prime}\left(I_{2} \oplus \mathrm{I}_{3}\right)$

## Priority Encoder

- In some practical cases more than one input can be high at a time, there we can not use simple encoder. In a priority encoder more than one input can be high at a time, A priority encoder is an encoder circuit that includes the priority function.
- The operation of the priority encoder is such that if two or more inputs are equal to 1 at the same time, the input having the highest priority will take precedence.

- They are often used to control interrupt requests by acting on the highest priority interrupt input. The priority encoders output corresponds to the currently active input which has the highest priority. So when an input with a higher priority is present, all other inputs with á lower priority will be ignored.
$I_{3}>I_{2}>I_{1}>I_{0}$
$\mathrm{O}_{0}=I_{3}+I_{3}{ }^{\prime} I_{2}^{\prime} I_{1}$
- $\mathrm{O}_{1}=\mathrm{I}_{2}+\mathrm{I}_{3}$

| $\mathrm{I}_{3}$ | $\mathrm{I}_{2}$ | $\mathrm{I}_{1}$ | $\mathrm{I}_{0}$ | $\mathrm{O}_{1}$ | $\mathrm{O}_{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 1 | 0 | 0 |
| 0 | 0 | 1 | d | 0 | 1 |
| 0 | 1 | d | d | 1 | 0 |
| 1 | d | d | d | 1 | 1 |

## SEQUENTIAL CIRCUITS

- Sequential Circuits consists of a combinational circuit to which memory elements are connected to form a feedback path. The memory elements are devices capable of storing binary information.
- The binary information stored in these elements at any given time defines the state of the sequential circuit at that time.



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- The sequential circuit receives binary information from external inputs $\left(x_{n}\right)$ that, together with the present state $\left(y_{n-1}\right)$ of the memory elements, determine the binary value of the outputs $\left(\mathrm{y}_{\mathrm{n}}\right)$.
- A sequential circuit is specified by a time sequence of inputs, outputs, and internal states.

$$
f\left(x_{n}, y_{n-1}\right)=y_{n}
$$


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## Latches

- Latch means to hold something or something which do not change.
- latches are the basic building blocks of any flip flop and they are capable of holding 1 bit until necessary.
- Storage elements that operate with signal levels are referred to as latches.
- Latches are level sensitive devices.


## NOR Latch

- The SR latch is a circuit with two cross-coupled NOR gates or two cross-coupled NAND gates, and two inputs labeled $S$ for set and $R$ for reset.
- Outputs $\mathrm{Q}_{n}$ and $\mathrm{Q}_{n}{ }^{\prime}$ are the complement of each other, in valid scenario.


| $S$ | $R$ | $Q_{n}$ | $Q_{n+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 1 |
| 0 | 0 | 1 |  |
| 0 | 1 | 0 |  |
| 0 | 1 | 1 |  |
| 1 | 0 | 0 |  |
| 1 | 0 | 1 |  |
| 1 | 1 | 0 |  |
| 1 | 1 | 1 |  |

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| $S$ | $R$ | $Q_{n}$ | $Q_{n+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | $X$ |
| 1 | 1 | 1 | $X$ |

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| $S$ | $R$ | $Q_{n}$ | $Q_{n+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | -1 |
| 0 | 0 | 1 |  |
| 0 | 1 | 0 |  |
| 0 | 1 | 1 |  |
| 1 | 0 | 0 |  |
| 1 | 0 | 1 |  |
| 1 | 1 | 0 |  |
| 1 | 1 | 1 |  |

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| $\mathbf{S}$ | $\mathbf{R}$ | $\mathbf{Q}_{n}$ | $\mathbf{Q}_{n+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | $\mathbf{x}$ |
| 0 | 0 | 1 | $x$ |
| 0 | 1 | 0 | 1 |
| 0 | 1 | 1 | 1 |
| 1 | 0 | 0 | 0 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 0 | 0 |
| 1 | 1 | 1 | 1 |

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## Notes

Latches are the basic circuits from which all flip-flops are constructed.
The storage elements (memory) used in clocked sequential circuits are called flipflops.
A flip-flop is a binary storage device capable of storing one bit of information. In a stable state, the output of a flip-flop is either 0 or 1 (it is called bi-stable multi-vibrator).

- A flip-flop is said to be stable if it has complementary behavior.
- Latches are level sensitive devices; flip-flops are edge-sensitive devices.
- Storage elements that operate with signallevels are referred to as latches; those controlled by a clock transition are flip-flops.


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## SR flip flop

1. The operation of the basic SR latch can be modified by providing an additional input signal that determines (controls), when the state of the latch can be changed by determining whether $S$ and $R$ (or $S$ and $R$ ) can affect the circuit.
2. It consists of the basic SR latch and two additional AND gates. The control input CP / $E_{n}$ acts as an enable signal for the other two inputs.
3. The outputs of the AND gates stay at the 0 as long as the enable signal remains at 0 as one input of AND gate gets 0 resulting in 0 as output. When the enable input goes to 1 , information from the $S$ or $R$ input is allowed to affect the latch.


Block Diagram


Implementation



Truth Table

| $\mathbf{s}$ | $\mathbf{R}$ | $\mathbf{Q}_{n}$ | $\mathbf{a}_{n+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 |  |
| 0 | 0 | 1 |  |
| 0 | 1 | 0 |  |
| 0 | 1 | 1 |  |
| 1 | 0 | 0 |  |
| 1 | 0 | 1 |  |
| 1 | 1 | 0 |  |
| 1 | 1 | 1 |  |

Truth Table

| $\mathbf{S}$ | $\mathbf{R}$ | $\mathbf{Q}_{\mathrm{n}}$ | $\mathbf{Q}_{\mathrm{n}+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | x |
| 1 | 1 | 1 | x |

K-Map

|  | $s r$ | $s^{\prime} r^{\prime}$ | $s^{\prime} r$ | $s r$ | $s^{\prime} r^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{q}$ |  | 00 | 01 | 11 | 10 |
| $q^{\prime}$ | 00 |  |  |  |  |
| $\mathbf{q}$ | 01 |  |  |  |  |

Characteristics Equation
$Q_{n+1}=$

Truth Table

| $\mathbf{S}$ | $\mathbf{R}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{n+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | X |
| 1 | 1 | 1 | x |

Characteristics Equation
Function Table

$$
Q_{n+1}=S+R^{\prime} Q_{n}
$$

Truth Table

| $\mathbf{S}$ | $\mathbf{R}$ | $\mathbf{Q}_{\mathrm{n}}$ | $\mathbf{Q}_{\mathrm{n}+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | $x$ |
| 1 | 1 | 1 | x |

Excitation Table


Truth Table

| $\mathbf{S}$ | $\mathbf{R}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{\mathrm{n}+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | $x$ |
| 1 | 1 | 1 | x |



| $\mathbf{S}$ | $\mathbf{R}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{\mathbf{n + 1}}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | $x$ |
| 1 | 1 | 1 | $x$ |



## Characteristics Equation

Function Table
Excitation Table

| $\mathbf{S}$ | $\mathbf{R}$ | $\mathbf{Q}_{n+1}$ |
| :---: | :---: | :---: |
| 0 | 0 | $\mathbf{Q}_{\mathbf{n}}$ |
| 0 | 1 | 0 |
| 1 | 0 | 1 |
| 1 | 1 | $X$ |$\quad$| $\mathbf{Q}_{n}$ | $\mathbf{Q}_{n+1}$ | $\mathbf{S}$ | $\mathbf{R}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | d |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 1 | d | 0 |


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Order of Priority -ve edge
+ve edge
-ve level
+ve level


## Notes

- Flip-flops can be either level-triggered (asynchronous, transparent or opaque) or edgetriggered (synchronous, or clocked).
- The term flip-flop has historically referred generically to both level-triggered and edgetriggered circuits that store a single bit of data using gates.
- Recently, some authors reserve the term flip-flop exclusively for discussing clocked circuits;
- the simple ones are commonly called transparent latches. Using this terminology, a levelsensitive flip-flop is called a transparent latch
- Whereas an edge-triggered flip-flop is simply called a flip-flop.
- Using either terminology, the term "flip-flop" refers to a device that stores a single bit of data. The terms "edge-triggered", and "level-triggered" may be used to avoid ambiguity.


## JK flip flop

- SR flip flop when both $S$ and $R=1$, results in invalid output. To resolve the problem we use JK Flip Flop. We take a feedback from the outputs.



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## JK flip flop

- SR flip flop when both $S$ and $R=1$, results in invalid output. To resolve the problem we use JK Flip Flop. We take a feedback from the outputs.


## Block Diagram



## Implementation



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Truth Table

| $J$ | $K$ | $Q_{n}$ | $Q_{n+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 |  |
| 0 | 0 | 1 |  |
| 0 | 1 | 0 |  |
| 0 | 1 | 1 |  |
| 1 | 0 | 0 |  |
| 1 | 0 | 1 |  |
| 1 | 1 | 0 |  |
| 1 | 1 | 1 |  |

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Truth Table

| $J$ | $K$ | $\mathbf{Q}_{n}$ | $\mathbf{Q}_{n+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | 1 |
| 1 | 1 | 1 | 0 |

K-Map

|  | $\mathbf{J K}$ | $\mathbf{J}^{\prime} \mathrm{K}^{\prime}$ | $\mathbf{J}^{\prime} \mathrm{K}$ | JK | JK |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{q}$ |  | 00 | 01 | 11 | 10 |
| $\mathbf{q}^{\prime}$ | 00 |  |  |  |  |
| $\mathbf{q}$ | 01 |  |  |  |  |

Characteristics Equation
$\mathrm{Q}_{\mathrm{n}+1}=$

Truth Table

| $\mathbf{J}$ | $\mathbf{K}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{\mathrm{n+1}}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | 1 |
| 1 | 1 | 1 | 0 |

Function Táble
Characteristics Equation

$$
Q_{n+1}=J Q_{n}^{\prime}+K^{\prime} Q_{n}
$$

Truth Table

| $J$ | $K$ | $\mathbf{Q}_{n}$ | $\mathbf{Q}_{n+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | 1 |
| 1 | 1 | 1 | 0 |

Excitation Table

| $\mathbf{Q}_{n}$ | $\mathbf{Q}_{n+1}$ | $\mathbf{b}$ | $-K$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 |  |  |
| 0 | 1 |  |  |
| 1 | 0 |  |  |
| 1 | 1 |  |  |

Truth Table

| $\mathbf{J}$ | K | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{\mathrm{n}+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | 1 |
| 1 | 1 | 1 | 0 |



## T(Togsle) flip flop

- The T (toggle) flip-flop is a complementing flip-flop and can be obtained from a JK flip-flop when inputs J and K are tied together.



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Truth Table

| $\mathbf{J}$ | K | $\mathbf{Q}_{\mathrm{n}}$ | $\mathbf{Q}_{\mathrm{n}+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | 1 |
| 1 | 1 | 1 | 0 |


| $T$ | $a_{n}$ | $a_{n+1}$ |
| :---: | :---: | :---: |
| 0 | 0 |  |
| 0 | $c_{1}$ |  |
| 1 | 0 |  |
| 1 | 1 |  |

Truth Table

| $\mathbf{T}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{\mathrm{n}+1}$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 0 |



Characteristics Equation
$Q_{n+1}=$

Truth Table

| $\mathbf{T}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{\mathrm{n}+1}$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 0 |

Function Table
Characteristics Equation


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Truth Table
Excitation Table

| $\mathbf{T}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{n+1}$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 0 |


| $Q_{n}$ | $Q_{n+1}$ | $T$ |
| :---: | :---: | :---: |
| 0 | 0 |  |
| 0 | 1 |  |
| 1 | 0 |  |
| 1 | 1 |  |

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Truth Table
State Diasram

| $\mathbf{T}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{\mathrm{n}+1}$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 0 |

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| Characteristics Equation | Function Table | Excitation Table | State Diagram |
| :---: | :---: | :---: | :---: |

$$
Q_{n+1}=T \oplus Q_{n} \bigcirc \begin{array}{|c|c|}
\hline 0 & Q_{n} \\
\hline 1 & Q_{n}^{\prime} \\
\hline
\end{array}
$$

| $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{\mathrm{n}+1}$ | $\mathbf{T}$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 0 |


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## D flip flop

- The D (Data/Delay) flip-flop, tracks the input at D and produces the same value as output.



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Truth Table

| $\mathbf{J}$ | K | $\mathbf{Q}_{\mathrm{n}}$ | $\mathbf{Q}_{\mathrm{n}+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | 1 |
| 1 | 1 | 1 | 0 |


| $D$ | $a_{n}$ | $a_{n+1}$ |
| :---: | :---: | :---: |
| 0 | 0 |  |
| 0 | $c_{1}$ |  |
| 1 | 0 |  |
| 1 | 1 |  |

Truth Table

| $\mathbf{D}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{\mathrm{n}+1}$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 0 |
| 1 | 0 | 1 |
| 1 | 1 | 1 |



Characteristics Equation
$Q_{n+1}=$

Truth Table

| $\mathbf{D}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{n+1}$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 0 |
| 1 | 0 | 1 |
| 1 | 1 | 1 |

Function Table
Characteristics Equation


Truth Table
Excitation Table

| $\mathbf{D}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{n+1}$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 0 |
| 1 | 0 | 1 |
| 1 | 1 | 1 |


| $Q_{n}$ | $Q_{n+1}$ | $D$ |
| :---: | :---: | :---: |
| 0 | 0 |  |
| 0 | 1 |  |
| 1 | 0 |  |
| 1 | 1 |  |

Truth Table
State Diasram

| $\mathbf{D}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{n+1}$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 0 |
| 1 | 0 | 1 |
| 1 | 1 | 1 |

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| Characteristics Equation | Function Table | Excitation Table | State Diagram |
| :--- | :--- | :--- | :--- |

$$
Q_{n+1}=D \bigcirc \begin{array}{|c|c|}
\hline D & a_{n+1} \\
\hline 0 & 0 \\
\hline 1 & 1 \\
\hline
\end{array}
$$

| $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{\mathrm{n}+1}$ | $\mathbf{D}$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 0 |
| 1 | 1 | 1 |


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## Flip Flops Conversion

1. We will require the Characteristics Table of target flip flop.
2. We will require the Excitation Table of given flip flop.
3. Determine the excitation values for characteristics table.
4. Obtain the expressions for input of given flip flop in terms of target.

Characteristics table of T- Flip Flop
Excitation Table of D flip flop


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| $\mathbf{T}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{n+1}$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 0 |


| $\mathbf{T}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{\mathrm{n}+1}$ | $\mathbf{D}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 |  |
| 0 | 1 | 1 |  |
| 1 | 0 | 1 |  |
| 1 | 1 | 0 |  |


| $Q_{n}$ | $Q_{n+1}$ | $D$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 0 |
| 1 | 1 | 1 |

## Convert D Flip Flop to T Flip Flop

Excitation values for characteristics table
Obtaining simplified expression

| $\mathbf{T}$ | $\mathbf{Q}_{n}$ | $\mathbf{Q}_{n+1}$ | $\mathbf{D}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 1 | 1 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | 0 |


| $T$ | $Q_{n}$ | $D$ |
| :---: | :---: | :---: |
| 0 | 0 |  |
| 0 | 1 |  |
| 1 | 0 |  |
| 1 | 1 |  |

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Obtaining simplified expression
Using K-Map to Simplify

| $\mathbf{T}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{D}$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 0 |


|  | $T$ | $T$ | $T$ |
| :---: | :---: | :---: | :---: |
| $q$ |  | 0 | 1 |
| $q^{\prime}$ | 0 |  |  |
| $q$ | 1 |  |  |

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## Convert D Flip Flop to T Flip Flop

## Using K-Map to Simplify

Block Diagram


$$
\mathrm{D}=\mathbf{Q}_{\mathrm{n}} \oplus \mathrm{~T}
$$

## Convert D Flip Flop to T Flip Flop

## Using K-Map to Simplify

Block Diagram


$$
\mathbf{D}=\mathbf{Q}_{\mathrm{n}} \oplus \mathrm{~T}
$$

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## Flip Flops Conversion

1. We will require the Characteristics Table of target flip flop.
2. We will require the Excitation Table of given flip flop.
3. Determine the excitation values for characteristics table.
4. Obtain the expressions for input of given flip flop in terms of target.

## Convert T flip flop to JK flip flop

Characteristics table of JK- Flip Flop

| $J$ | $K$ | $Q_{n}$ | $\mathbf{Q}_{n+1}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 |  |
| 0 | 0 | 1 |  |
| 0 | 1 | 0 |  |
| 0 | 1 | 1 |  |
| 1 | 0 | 0 |  |
| 1 | 0 | 1 |  |
| 1 | 1 | 0 |  |
| 1 | 1 | 1 |  |

Excitation Table of T flip flop

| $Q_{n}$ | $Q_{n+1}$ | $\mathbf{T}$ |
| :---: | :---: | :---: |
| 0 | 0 |  |
| 0 | 1 |  |
| 1 | 0 |  |
| 1 | 1 |  |


| J | K | $\mathrm{Q}_{n}$ | $\mathrm{Q}_{n+1}$ |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |  |  |  |  |  |
| 0 | 0 | 1 | 1 | J | K | $\mathrm{Q}_{\mathrm{n}}$ | $Q_{n+1}$ | T |
| 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |  |
| 0 | 1 | 1 | 0 | 0 | 0 | 1 | 1 | 2 |
| 1 | 0 | 0 | 1 | 0 | 1 | 0 | 0 | , |
| 1 | 0 | 1 | 1 | 0 | 1 | 1 | 0 |  |
| 1 | 1 | 0 | 1 | 1 | 0 | 0 | 1 |  |
| 1 | 1 | 1 | 0 | 1 | 0 | 1 | 1 |  |
|  |  |  |  | 1 | 1 | 0 | 1 |  |
|  |  |  |  | 1 | 1 | 1 | 0 |  |


| $Q_{n}$ | $Q_{n+1}$ | $T$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 0 |
| 1 |  |  |

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## Convert T flip flop to JK flip flop

Excitation values for characteristics table

| $\mathbf{J}$ | $\mathbf{K}$ | $\mathbf{Q}_{\mathbf{n}}$ | $\mathbf{Q}_{n+1}$ | $\mathbf{T}$ |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 | 0 |
| 0 | 1 | 0 | 0 | 0 |
| 0 | 1 | 1 | 0 | 1 |
| 1 | 0 | 0 | 1 | 1 |
| 1 | 0 | 1 | 1 | 0 |
| 1 | 1 | 0 | 1 | 1 |
| 1 | 1 | 1 | 0 | 1 |

## Convert T flip flop to JK flip flop

## Using K-Map to Simplify

| $J$ | $K$ | $\mathbf{Q}_{\mathrm{n}}$ | $\mathbf{T}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 0 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 1 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 0 | 1 |
| 1 | 1 | 1 | 1 |


|  | JK | $J^{\prime} K^{\prime}$ | JK | JK | JK |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{q}$ |  | 00 | 01 | 11 | 10 |
| $\mathbf{q}^{\prime}$ | 00 |  |  |  |  |
| $\mathbf{q}$ | 01 |  |  |  |  |

$T=$

## Convert T flip flop to JK flip flop

## Block Diagram

$\mathrm{T}=$


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## Convert T flip flop to JK flip flop

## Using K-Map to Simplify

Block Diagram


$$
\begin{gathered}
\mathbf{T}=\mathbf{J} \mathbf{Q}_{n}^{\prime}+\mathbf{K} \mathbf{Q}_{\mathrm{n}} \\
\text { www.knowledgegate.in }
\end{gathered}
$$

## Flip Flops Conversion

1. We will require the Characteristics Table of target flip flop.
2. We will require the Excitation Table of given flip flop.
3. Determine the excitation values for characteristics table.
4. Obtain the expressions for input of given flip flop in terms of target.

## Basics of counters

- A counter is a sequential circuit that goes through a predetermined sequence of binary states upon the application of input pulses. A counter that follows the binary number sequence is called a binary counter.
- An $n$ - bit binary counter consists of $n$ flip-flops and can count in binary from 0 through $2^{n}-1$. Counters are available in two categories: synchronous counters and Ripple counters (asynchronous).


Instruction register (IR)


VVWVN.INIUVVIEU,






Q Find out the counting sequence for the counter below?
The characteristics equation for JK Flip Flop is $Q_{n+1}=$ $\mathrm{Q}_{0 \mathrm{~N}}=$

The characteristics equation for T Flip Flop is $Q_{n+1}=$ $\mathrm{Q}_{1 \mathrm{~N}}=$


| Present State |  | Next State |  |
| :---: | :---: | :---: | :---: |
| $\mathrm{Q}_{1 \mathrm{p}}$ | $\mathrm{Q}_{\mathrm{op}}$ | $\mathrm{Q}_{1 \mathrm{~N}}$ | $\mathrm{O}_{\mathrm{oN}}$ |
| 0 | 0 |  |  |
| 0 | 1 |  |  |
| 1 | 0 |  |  |
| 1 | 1 |  |  |

Q Consider a 3-bit counter, designed using T flip-flops, as shown below:


| Present State |  |  | Next State |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{Q}_{20}$ | $\mathrm{Q}_{10}$ | $\mathrm{Q}_{00}$ | $\mathrm{Q}_{2 N}$ | $\mathrm{Q}_{1 N}$ | $\mathrm{Q}_{0 N}$ |
| 0 | 0 | 0 |  |  |  |
| 0 | 0 | 1 |  |  |  |
| 0 | 1 | 0 |  |  |  |
| 0 | 1 | 1 |  |  |  |
| 1 | 0 | 0 |  |  |  |
| 1 | 0 | 1 |  |  |  |
| 1 | 1 | 0 |  |  |  |
| 1 | 1 | 1 |  |  |  |

Q Design a synchronous counter for sequence: $0 \rightarrow 1 \rightarrow 2 \rightarrow 3 \rightarrow 0$, using $T$ flip flop.

| Present State |  | Next State |  |
| :---: | :---: | :---: | :---: |
| $\mathrm{Q}_{1 \mathrm{p}}$ | $\mathrm{Q}_{0 \mathrm{p}}$ | $\mathrm{Q}_{1 \mathrm{~N}}$ | $\mathrm{Q}_{0 \mathrm{~N}}$ |
| 0 | 0 |  |  |
| 0 | 1 |  |  |
| 1 | 0 |  |  |
| 1 | 1 |  |  |

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Q Example: Design a synchronous counter for sequence: $0 \rightarrow 1 \rightarrow 2 \rightarrow 3 \rightarrow 0$, using D flip flop.

| Present State |  | Next State |  |
| :---: | :---: | :---: | :---: |
| $\mathrm{Q}_{1 \mathrm{p}}$ | $\mathrm{Q}_{\mathrm{op}}$ | $\mathrm{Q}_{1 \mathrm{~N}}$ | $\mathrm{Q}_{\mathrm{oN}}$ |
| 0 | 0 |  |  |
| 0 | 1 |  |  |
| 1 | 0 |  |  |
| 1 | 1 |  |  |

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Q Design synchronous counter for sequence: $0 \rightarrow 1 \rightarrow 3 \rightarrow 4 \rightarrow 5 \rightarrow 7 \rightarrow 0$, using $T$ flip-flop.

| Present State |  |  | Next State |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{Q}_{2 \mathrm{p}}$ | $\mathrm{Q}_{1 \mathrm{p}}$ | $\mathrm{Q}_{0 \mathrm{p}}$ | $\mathrm{Q}_{2 \mathrm{~N}}$ | $\mathrm{Q}_{1 \mathrm{~N}}$ | $\mathrm{Q}_{0 \mathrm{~N}}$ |
| 0 | 0 | 0 |  |  |  |
| 0 | 0 | 1 |  |  |  |
| 0 | 1 | 0 |  |  |  |
| 0 | 1 | 1 |  |  |  |
| 1 | 0 | 0 |  |  |  |
| 1 | 0 | 1 |  |  |  |
| 1 | 1 | 0 |  |  |  |
| 1 | 1 | 1 |  |  |  |

## RIPPLE COUNTERS

1. A binary ripple counter consists of a series connection of complementing flip-flops, with the output of each flip-flop connected to the input of the next higher order flip-flop.
2. The flip-flop holding the least significant bit receives the incoming count pulses.

$\mathbf{Q}$ In the sequential circuit shown below, if the initial value of the output $\mathrm{Q}_{1} \mathrm{Q}_{0}$ is 00 , what are the next four values of $\mathrm{Q}_{1} \mathrm{Q}_{0}$ ?


| Present State |  | Next State |  |
| :---: | :---: | :---: | :---: |
| $\mathrm{Q}_{1 \mathrm{p}}$ | $\mathrm{Q}_{0 p}$ | $\mathrm{Q}_{1 \mathrm{~N}}$ | $\mathrm{Q}_{\mathrm{ON}}$ |
| 0 | 0 |  |  |
| 0 | 1 |  |  |
| 1 | 0 |  |  |
| 1 | 1 |  |  |

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| Nature of Clock | Nature of Feedback | Nature of counting |
| :---: | :---: | :---: |
| + ve |  | UP Counting |
| +ve |  | Down Counting |
| -ve |  | UP Counting |
| -ve |  | Down Counting |

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## Asynchronous counters

| SYNCHRONOUS COUNTER | ASYNCHRONOUS COUNTER |
| :---: | :---: |
| In synchronous counter, all flip flops are triggered with same clock simultaneously. | In asynchronous counter, different flip flops are triggered with different clock, not simultaneously. |
| Synchronous Counter is faster than asynchronous counter in operation. $\mathrm{T}_{\text {delay }}=\mathrm{T}_{\text {FF }}+\mathrm{T}_{\mathrm{Cc}}$ | Asynchronous Counter is slower than synchronous counter in operation. $T_{\text {delay }}=n \times T_{\text {FF }}+T \mathrm{Cc}$ |
| Synchronous Counter is also called Parallel Counter. | Asynchronous Counter is also called Serial Counter. |
| Synchronous Counter designing as well implementation are complex due to increasing the number of states. | Asynchronous Counter designing as well as implementation is very easy. |
| Synchronous Counter will operate in any desired count sequence. | Asynchronous Counter will operate only in fixed count sequence (UP/DOWN). |
| Synchronous Counter examples are: Ring counter, Johnson counter. | Asynchronous Counter examples are: Ripple UP counter, Ripple DOWN counter. |
| A synchronous sequential circuit is a system whose behavior can be defined from the knowledge of its signals at discrete instants of time. | The behavior of an asynchronous sequential circuit depends upon the input signals at any instant of time and the order in which the inputs change. |

## Resisters

- Registers are basically storing devices which are also designed using basic element called flip-flop.
- D-flip-flop are most popular choice for register because they don't perform any functionality and output is simply based on current inputso, they act as a buffer.
- Apart from storing registers sometimes also be used in performing basic mathematical operation like multiply by 2 by left shit and divide by 2 by right shift.


## Shift Register Types \& Operations

- There are four different modes in which registers operate.

1. Serial In-Serial Out (SISO)
2. Serial In-Parallel Out (SIPO)
3. Parallel In-Serial Out (PISO)
4. Parallel In- Parallel Out (PIPO)

## Serial In-Serial Out (SISO)

- The shift register, which allows serial input (one bit after the other through a single data line) and produces a serial output is known as Serial-In Serial-Out shift register.
- Since there is only one output, the data leaves the shift register one bit at a time in a serial pattern, thus the name Serial-In Serial-Out Shift Register.



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| Sequence | Input | $\mathrm{Q}_{2}$ | $\mathrm{Q}_{1}$ | $\mathrm{Q}_{0}$ | Output |
| :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |

- The main use of a SISO is to act as a delay element.
- In SISO registers to provide n bit data serially in it requires n clock pulse and to provide serial output it requires $\mathrm{n}-1$ clock pulses.

|  | No of clock <br> (write) | No of clock <br> (Read) | total |
| :---: | :---: | :---: | :---: |
| SISO |  |  |  |
| SIPO |  |  |  |
| PISO |  |  |  |
| PIPO |  |  |  |

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## Serial-In Parallel-Out shift Register (SIPO)

- The shift register, which allows serial input (one bit after the other through a single data line) and produces a parallel output is known as Serial-In Parallel-Out shift register.



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- They are used in communication lines where demultiplexing of a data line into several parallel lines is required because the main use of the SIPO register is to convert serial data into parallel data.

| SISO | No of clock <br> (write) | No of clock <br> (Read) | total |
| :---: | :---: | :---: | :---: |
| SIPO |  | $\mathrm{n}-1$ | $2 \mathrm{n}-1$ |
| PISO |  |  |  |
| PIPO |  |  |  |

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## Parallel-In Serial-Out Shift Register (PISO)

- The shift register, which allows parallel input (data is given separately to each flip flop and in a simultaneous manner) and produces a serial output is known as Parallel-In Serial-Out shift register.


- The circuit consists of four D flip-flops which are connected.
- The clock input is directly connected to all the flip flops but the input data is connected individually to each flip flop through a multiplexer at the input of every flip flop.
- The output of the previous flip flop and parallel data input are connected to the input of the MUX and the output of MUX is connected to the next flip flop.
- A Parallel in Serial out (PISO) shift register us used to convert parallel data to serial data.

| SISO | No of clock <br> (write) | No of clock <br> (Read) | total |
| :---: | :---: | :---: | :---: |
| SIPO | n | $\mathrm{n}-1$ | $2 \mathrm{n}-1$ |
| PISO |  | 0 | n |
| PIPO |  |  |  |

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## Parallel-In Parallel-Out Shift Register (PIPO)

- The shift register, which allows parallel input (data is given separately to each flip flop and in a simultaneous manner) and also produces a parallel output is known as Parallel-In parallel-Out shift register.



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- The circuit consists of four D flip-flops which are connected.
- In this type of register, there are no interconnections between the individual flip-flops since no serial shifting of the data is required.
- Data is given as input separately for each flip flop and in the same way, outputalso collected individually from each flip flop.
- A Parallel in Parallel out (PIPO) shift register is used as a temporary storage device and like SISO Shift register it acts as a delay element.
- For parallel input it requires 1 clock pulse and for paraliel output it requires 0 clock.


| SISO | No of clock <br> (write) | No of clock <br> (Read) | total |
| :---: | :---: | :---: | :---: |
| SIPO | n | $\mathrm{n}-1$ | $2 \mathrm{n}-1$ |
| PISO | 1 | 0 | n |
| PIPO |  | $\mathrm{n}-1$ | n |

## 4-Bit Ring Counter

- A ring counter is a circular shift register with only one flip-flop being set at any particular time; all others are cleared.
- The single bit is shifted from one flip-flop to the next to produce the sequence of timing signals.
- Output of the last flip-flop is connected to the input of the first flip-flop in case of ring counter.
- No. of states in Ring counter = No. of flip-flop used



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- Number of unused states in Ring Counter is $2^{2 n}-n$


| CLK | $Q_{3}$ | $Q_{2}$ | $Q_{1}$ | $Q_{0}$ |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 1 | 0 | 0 | 0 |
| 1 | 0 | 1 | 0 | 0 |
| 2 | 0 | 0 | 1 | 0 |
| 3 | 0 | 0 | 0 | 1 |
| 0 | 1 | 0 | 0 | 0 |

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- A k-bit ring counter circulates a single bit among the flip-flops to provide $k$ distinguishable states.
- The number of states can be doubled if the shift register is connected as a switch-tail ring counter. A switch-tail ring counter is a circular shift register with the complemented output of the last flip-flop connected to the input of the first flip-flop.



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| CLK | $\mathrm{Q}_{3}$ | Q, | $Q_{1}$ | Q |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 |
| 1 | 1 | 0 | 0 | 0 |
| 2 |  | 1 | 0 | 0 |
| 3 | 1 | 1 | 1 | 0 |
| 4 | 1 | 1 | 1 | 1 |
| 5 | 0 | 1 | 1 | 1 |
| 6 | 0 | 0 | 1 | 1 |
| 7 | 0 | 0 | 0 | 1 |
| 0 | 0 | 0 | 0 | 0 |
| WWVW.KIIOWIEQ\%E\%dLE.II |  |  |  |  |

## Basics

- Main idea in number system is counting and representation of quantity
- In stone age or even today the basic idea of counting in Unary counting
- Remember how we started counting on fingers and using abacus
- How can we use it for counting sheep or anything?



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- But when we want to count larger quantity in anything, day to day life, business, maths or research, we cannot work with unary system.

S. msung pays Apple $\$ 1$ Billion sending 30 trucks full of 5 cent coins


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- So next popular system we use in real life is decimal system, may be because we have 10 fingers on over hands, in history different cultures have been using base 10 for general purpose counting, Indian, British, roman, Arabic etc.

- Thought Aryabhata (3500 years back) was the first one known to extensively worked on the idea of zero, pie, number system, trigonometry, quadratic equations, astronomy etc.
- In Indian culture we were working on very complex math from much early time, Aryabhata was may be first one, who have properly written and published his understanding, and some



| Number System | Base or Radix | Digits or symbols |
| :---: | :---: | :--- |
| Binary | 2 | 0,1 |
| Octal | 8 | $0,1,2,3,4,5,6,7$ |
| Hexadecimal | 16 | $0,1,2,3,4,5,6,7,8,9, A, B, C, D, E, F$ |


| Number System | Base or Radix | Digits or symbols |
| :---: | :---: | :---: |
| Unary | 1 | $0 / 1$ |
| Decimal | 10 | $0,1,2,3,4,5,6,7,8,9$ |

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## Decimal Number System

- The Decimal Number System uses 10 different digits: $0,1,2,3,4,5,6,7,8,9$. It's called.a "base-10" system because it relies on these 10 digits and uses powers of 10 to form numbers. This is the number system we commonly use in everyday activities.
- Example: $(739.2)_{10}$, where 739.2 is a shorthand notation for what should be written as



## Binary Number System

- In the Binary Number System, you only use two digits: 0 and 1. Each digit in a binary number is multiplied by a power of 2 (like $2^{0}, 2^{1}, 2^{2}$, and so on). Then you add these products together to get the number in decimal form.

- Similarly, there can be many bases and they can be easily converted to decimal number system.
- An example of a base- 5 number is


Convert from Decimal to Any Base
$(4213.21)_{5} \longrightarrow(558.44)_{10} \longrightarrow(1425.3036)_{7}^{7}$

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## $(83)_{10} \longrightarrow(1010011)_{2}$

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## $(109.125)_{10} \longrightarrow(1101101.001)_{2}$

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## Octal Number System



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## Hexadecimal Number System

- When the base of a number system is higher than 10 , we use letters from the alphabet to represent the extra digits. For instance, in the Hexadecimal System, which is base-16, the first 10 digits are the same as in the Decimal System: 0 to 9 . After that, the letters $A, B, C, D, E$, and F stand for the numbers $10,11,12,13,14$, and 15 respectively. The hexadecimal system is most popular in digital systems.

- In case a binary number usually becomes so large so better idea is to club digits in a group of three or four, leading to octal, hexadecimal

| Hexadecimal System / Base 16 System |  |
| :---: | :---: |
| 0000 | 0 |
| 0001 | 1 |
| 0010 | 2 |
| 0011 | 3 |
| 0100 | 4 |
| 0101 | 5 |
| 0110 | 6 |
| 0111 | 7 |
| 1000 | 8 |
| 1001 | 9 |
| 1010 | $10-\mathrm{A}$ |
| 1011 | $11-\mathrm{B}$ |
| 1100 | $12-\mathrm{C}$ |
| 1101 | $13-\mathrm{D}$ |
| 1110 | $14-\mathrm{E}$ |
| 1111 | $15-\mathrm{F}$ |
|  |  |


| Octal System / Base 8 System |  |
| :---: | :---: |
| 000 | 0 |
| 001 | 1 |
| 010 | 2 |
| 011 | 3 |
| 100 | 4 |
| 101 | 5 |
| 110 | 6 |
| 111 | 7 |



## $(110010.111)_{2}$ $(—$ _-_) 4

( 110010.111 $)_{2}$
(른) $)_{8}$
$(110010.111)_{2}$

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## $(62.7)_{8}$


$(\ldots .)_{16}$

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## Representation of a number

Number


Signed
number

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## Unsigned number

- Until now, we have considered only unsigned numbers. Unsigned numbers without any positive or negative sign, these numbers represent only magnitude. In real life, we have only unsigned numbers (considered as positive) (natural numbers) for e.g. number of stars, number of trees etc.
- If $n$ bits are used to store the number, then all the $n$ bits are used to store number / magnitude / absolute value. If we have a $n$ bit Unsigned number than range is from 0 to $2^{n}-1$


## Signed number

- For modern-day math tasks like calculating profit and loss or measuring temperature, we often need to use negative numbers. In daily life, we write these numbers with both their sign and magnitude. For example, +9 shows a positive number and -543 indicates a negative one. We use the symbols "+" and "-" to show if the number is positive or negative, respectively.



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- In a computer, an n-bit word can store a signed binary number. One usually the leftmost bit is set aside to show if the number is positive or negative, and the other $n-1$ bits represent the size or magnitude of the number.
- Because of the limitation of the computer hardware as circuit only understand 0 and 1 , every piece of information in computer must be represented in terms of numbers i.e. 0 and 1 , weather it is sign or magnitude. where first left most bit 0 means + Ve number and 1 means Ve number.


Signed number
representation

Signed magnitude convention(here the negative number is represented by it's sign)


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| Decimal | Signed Magnitude | 1's Complement | 2's Complement |
| :---: | :---: | :---: | :---: |
| +7 | 0111 | 0111 | 0111 |
| +6 | 0110 | 0110 | $0110 \sim$ |
| +5 | 0101 | 0101 | - 0101 |
| +4 | 0100 | 0100 | 0100 |
| +3 | 0011 | 0011 | 0011 |
| +2 | 0010 | $0010 \square$ | 0010 |
| +1 | 0001 | 0001 | 0001 |
| +0 | 0000 | 0000 | 0000 |
| -0 | 1000 | 1111 | 0000 |
| -1 | 1001 - | 1110 | 1111 |
| -2 | 1010 | 1101 | 1110 |
| -3 | 1011 | 1100 | 1101 |
| -4 | 1100 | 1011 | 1100 |
| -5 | 1101 | 1010 | 1011 |
| -6 | 1110 | 1001 | 1010 |
| -7 | 1111 M N Kn | MMPd91000ate in | 1001 |
| -8 | Hine- |  | 1000 |

- Represent decimal number -13 in all three methods?
- Signed Magnitude
- 1's Complement
- 2's Complement
- Read number 10101 according to all three representation?
- Signed Magnitude
- 1's Complement
- 2's Complement


## Sign magnitude representation

- Introduction: - It is the simplest form of representation, where in an n-bit word, the right mostn-1 bits (from LSB) hold the magnitude of the number and $\mathrm{n}^{\text {th }}$ bit (Left most, MSB) is assigned for sigh, where $0 \rightarrow$ +ve and $1 \rightarrow$-ve.
- Range: - from $-\left(2^{n-1}-1\right)$ to $+\left(2^{n-1}-1\right)$ total $2^{n}-1$
- Advantage: - Easy to read and write, because it is a weighted code system, directly reading and writing is possible.
- Disadvantage: - One drawback in sign-magnitude representation is addition and subtraction require separate handling of the sign and the magnitude. One more problem is there are two representation for 0 , i.e. $-0 \&+0$, because of which there is a loss of one presentation, and create confusion.
- Number extension : - In computer the size of a location is fixed, so it is often required to write a smaller number in a larger space, this can be done by sign extension, where, the magnitude is written as far right as possible, sign bit is written left most, and all the empty cells must be filled/padded with zero.


## 1's Complement Representation

- Introduction: - here the negative number is represented by taking complement.
- Range: - from $-\left(2^{n-1}-1\right)$ to $+\left(2^{n-1}-1\right)$, total $2^{n}-1$ representation
- Advantage: - Easy to calculate, used as logical operations.
- Disadvantage: - One drawback in 1's complement representation is addition and subtraction in difficult, Not a weighted code system. there are two representation for 0 , i.e. $-0 \&+0$, because of which there is a loss of one presentation. Neither we can write a negative number directly nor we can read a negative number directly.
- Number extension: In computer the size of a location is fixed, so it is often required to write a smaller number in a larger space, this can be done by sign extension, where, if the number is positive extension is done same as that of sign magnitude extension. If the number is -ve then we must write the number as far right as possible and then all the empty cells must be filled with 1 (in negative logic 0 is represented by 1 ).


## 2's Complement Representation

- Introduction: - if 1 is added to 1's complement of a binary number, the resulting number known as the 2's complement of the binary number.
- Range: - from $-\left(2^{n-1}\right)$ to $+\left(2^{n-1}-1\right)$, total $2^{n}$ representation


## Advantage: -

- Easy to do arithmetic operations with 2's complement representation. As end round carry can be discarded. Has only one representation for zero which is always positive, therefore provides better clarity, efficiency and more range with no wastage. It is a waited code system, direct reading and writing is possible.
- Disadvantage: - One drawback in 2's complement representation is relatively difficult to understand. But in application like floating point representation it is not much required, and it takes more space.
- Number extension: In computer the size of a location is fixed, so it is often required to write a smaller number in a larger space, this can be done by sign extension, where, if the number is positive extension is done same as that of sign magnitude extension. If the number is -ve then we must write the number as far right as possible and then all the empty cells must be filled with 1(in negative logic 0 is represented by 1).
- Subtraction using 2's Complement arithmetic
- 9-4 =
- 7-11 =


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## Gray Code

- Gray Code is a binary numbering system where two successive numbers differ in only one bit. It's essential because it reduces errors in digital systems, particularly in analog-to-digital conversions.
- Applications include rotary encoders in mechanical systems and minimizing errors in multi-bit data transfer. Gray Code contributes to system reliability by ensuring that only one bit changes at a time, reducing the risk of transitional errors.
- It's particularly useful in situations where an error in multiple bits changing at once could be catastrophic, like in control systems.


## 11001110

$\mathrm{b}_{0} 0$

| - $\mathrm{G}_{0}=\mathrm{B}_{1} \oplus \mathrm{~B}_{0}$ |  |  |  |  |  | Hbitay |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - $\mathrm{G}_{0}=\mathrm{B}_{1} \oplus \mathrm{~B}_{0}$ | $\frac{\mathrm{B}_{\mathrm{t}}}{0}$ | ${ }_{0}^{\text {Bi }}$ | ${ }_{0}^{\text {B }}$ | ${ }^{B_{1}}$ |  | $G_{i}$ $G_{j}$ <br> 0 0 | $\mathrm{G}_{2}$ |  |
| - $\mathrm{G}_{1}=\mathrm{B}_{2} \oplus \mathrm{~B}_{1}$ | 0 | 0 | 0 | 1 | 0 | 0.0 |  |  |
| - $\mathrm{G}_{2}=\mathrm{B}_{3} \oplus \mathrm{~B}_{2}$ |  | 0 | 1 | 0 |  | a 0 |  |  |
|  |  | 1 | 0 |  | 5 | 01 | 1 | 0 |
| - $\mathrm{G}_{3}=\mathrm{B}_{3}$ |  | 1 |  |  |  | 0 | 1 | 1 |
|  |  | 1 |  |  |  | $0 \quad 1$ | 0 | 1 |
|  |  | \% | 0 | 1 |  | 0 1 1 | $\bigcirc$ | 0 |
|  |  | 0 | 0 | 1 |  | 1.1 | 0 | 1 |
|  | 1 | 0 | 1 | 0 |  | 1 | 1 | 1 |
|  | 1 | 0 | 1 |  |  | 11 | 1 | 0 |
|  | , | 1 | 0 | 0 |  | 10 | 1 | 0 |
|  | 1 | 1 | 1 | 1 | 1 | 10 | 1 | 1 |
| Lw | 1 | 1 | 1 | 1 | $\frac{1}{1}$ | 10 | 0 | 0 |

## 11001110 <br> 

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- $\mathrm{B}_{0}=\mathrm{B}_{1} \bigoplus \mathrm{G}_{0}$ $B_{1}=B_{2} \bigoplus G_{1}$ $B_{2}=B_{3} \bigoplus G_{2}$ $B_{3}=G_{3}$

| $\mathbf{G 3}$ | $\mathbf{G} 2$ | $\mathbf{G} \mathbf{G 0}$ | $\mathbf{B 3}$ | $\mathbf{B 2}$ | $\mathbf{B} \mathbf{B 0}$ | $\mathbf{B 0}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 0 | 1 | 0 | 0 | 0 | 1 |
| 0 | 0 | 1 | 1 | 0 | 0 | 1 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 1 | 1 |
| 0 | 1 | 1 | 0 | 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 1 | 0 | 1 | 0 | 1 |
| 0 | 1 | 0 | 1 | 0 | 1 | 1 | 0 |
| 0 | 1 | 0 | 0 | 0 | 1 | 1 | 1 |
| 1 | 1 | 0 | 0 | 1 | 0 | 0 | 0 |
| 1 | 1 | 0 | 1 | 1 | 0 | 0 | 1 |
| 1 | 1 | 1 | 1 | 1 | 0 | 1 | 0 |
| 1 | 1 | 1 | 0 | 1 | 0 | 1 | 1 |
| 1 | 0 | 1 | 0 | 1 | 1 | 0 | 0 |
| 1 | 0 | 1 | 1 | 1 | 1 | 0 | 1 |
| 1 | 0 | 0 | 1 | 1 | 1 | 1 | 0 |
| $1<$ | 0 | 0 | 0 | 1 | 1 | 1 | 1 |

## Binary-Coded Decimal Code (BCD)

- BCD (Binary-Coded Decimal) is a class of binary encodings where each decimal digit is represented by a fixed number of binary digits, usually four. It's needed because it simplifies the process of converting between binary and decimal numbers, making calculations more accurate and easier to check.
- Applications of BCD are widespread, including in digital clocks, calculators, and financial systems where decimal representation is critical for accuracy. It is also useful in systems that require easy conversion to human-readable decimal numbers.
- BCD contributes to computational efficiency and accuracy. It makes it simpler to convert between binary and decimal, reducing the chance of errors that could occur in calculations or data transfers.

| Decimal | $\mathrm{B}_{3}$ | $\mathrm{~B}_{2}$ | $\mathrm{~B}_{1}$ | $\mathrm{~B}_{0}$ | $\mathrm{D}_{3}$ | $\mathrm{D}_{2}$ | $\mathrm{D}_{1}$ | $\mathrm{D}_{0}$ | $\mathrm{D}_{3}$ | $\mathrm{D}_{2}$ | $\mathrm{D}_{1}$ | $\mathrm{D}_{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |
| 2 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 |
| 3 | 0 | 0 | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 |
| 4 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| 5 | 0 | 1 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 1 |
| 6 | 0 | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 0 |
| 7 | 0 | 1 | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 |
| 8 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 |
| 9 | 1 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 1 |
| 10 | 1 | 0 | 1 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 |
| 11 | 1 | 0 | 1 | 1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 1 |
| 12 | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 1 | 0 |
| 13 | 1 | 1 | 0 | 1 | 0 | 0 | 0 | 1 | 0 | 0 | 1 | 1 |
| 14 | 1 | 1 | 1 | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 0 | 0 |
| 15 | 1 | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

- Represent the unsigned decimal number 965 and 672 in BCD and then show the steps necessary to find their sum?

| 965 | 1001 | 0110 | 0101 |
| :--- | :--- | :--- | :--- |
| 672 | 0110 | 0111 | 0010 |
| FD7 | 1111 | 1101 | 0111 |

\(\left.\begin{array}{cccc}FD7 \& \begin{array}{c}1111 <br>

+0110\end{array} \& 1101 \& 0110\end{array}\right]\)| 0111 |
| :---: |
| 0000 |
| $\frac{0110}{15}$ |

If sum $>9$ then add 6.for correctionledgegate. in


## Excess-3 Code

- Excess-3 Code is a binary-coded decimal (BCD) system that represents each digit of a decimal number by its corresponding 4-bit binary representation, but adds an offset of 3 to each digit. This makes arithmetic operations like addition and subtraction easier to implement.
- Excess-3 Code is primarily used in older computer systems and some types of calculators. It was important for minimizing hardware in early digital systems, where computational resources were limited.
- The main contribution of Excess-3 Code is that it simplifies the hardware required for arithmetic operations. Its self-complementing feature helps in carrying out subtraction by addition, thereby reducing computational complexity.

| Decimal | Excess-3 |
| :---: | :---: |
| 0 | 0011 |
| 1 | 0100 |
| 2 | 0101 |
| 3 | 0110 |
| 4 | 0111 |
| 5 | 1000 |
| 6 | 1001 |
| 7 | 1010 |
| 8 | 1011 |
| 9 | 1100 |

- Example: 395 in decimal will be represented in excess-3 code as:
011011001000.


## 395

- 3 in excess-3 representation is: $3+3=6$ (0110)
- 9 in excess-3 representation is: $9+3=12$ (1100)
- 5 in excess-3 representation is: $5+3=8$ (1000)
- First, convert the decimal digits to their 4-bit Excess-3 codes:
- 5 in Excess-3: 5+3=8 in decimal, which is 1000 in binary.
- 7 in Excess-3: $7+3=10$ in decimal, which is 1010 in binary.
- Add the Excess-3 coded numbers:
- $1000+1010=10010$ in binary, which is 18 in decimal.
- Now, subtract 3 from the sum to remove the excess and get the real sum:
$18-3=15$
15 in Excess-3 code, real sum is 12

